

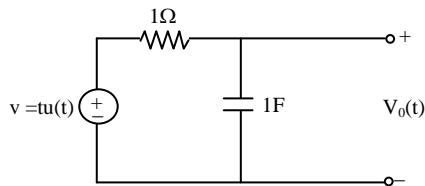


08. Ans: (d)

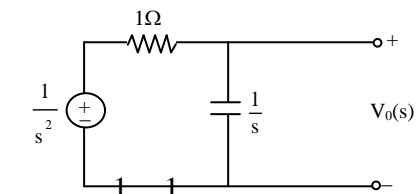
Sol: In case of Michelson interferometer if we use polychromatic light the pattern is not readable & no more useful, also color fringes overlap on each other.

09. Ans: (b)

Sol:



Converting into Laplace domain we get



$$V_o(s) = \frac{\frac{1}{s^2} \times \frac{1}{s^2}}{1 + \frac{1}{s}} = \frac{1}{s^3 \left(\frac{s+1}{s} \right)}$$

$$= \frac{1}{s^2(s+1)}$$

$$V_o(s) = \frac{A}{s} + \frac{B}{s^2} + \frac{C}{s+1}$$

$$B = \frac{1}{s+1} \Big|_{s=0} = 1$$

$$A = \frac{dB}{ds} \Big|_{s=0} = \frac{(s+1) \times 0 - 1}{(s+1)^2} \Big|_{s=0} = -1$$

$$C = \frac{1}{s^2} \Big|_{s=-1} = 1$$

$$V_o(s) = \frac{-1}{s} + \frac{1}{s^2} + \frac{1}{s+1}$$

Apply inverse Laplace transform on both sides

$$V_o(t) = -u(t) + tu(t) + e^{-t} u(t)$$

$$V_o(t) = ((t-1) + e^{-t})u(t), t > 0$$

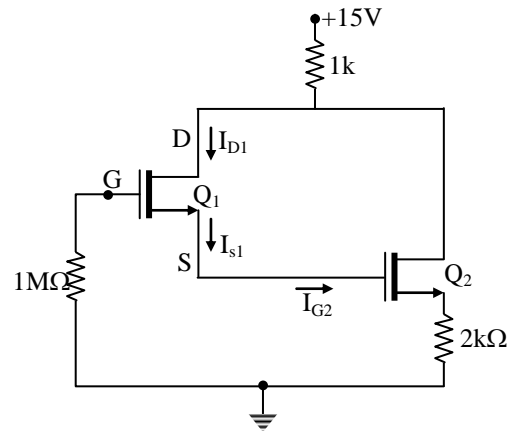
10. Ans: (d)

Sol: If $V_A > V_B$ all diodes are forward biased hence $R_{AB} = 0$. If $V_A < V_B$, D_1 D_3 RB & D_2 FB

$$\therefore R_{AB} = 36 \Omega + 18 \Omega = 54 \Omega$$

11. Ans: -4 (No range)

Sol:



Step(1): From the circuit,

$$I_{S1} = I_{G2} = 0 \quad (1) \quad [\because I_G = 0 \text{ in MOSFETs}]$$

$$\Rightarrow I_{D1} = I_{S1} = 0 \quad (2) \quad [\because I_D = I_S \text{ in MOSFETs}]$$

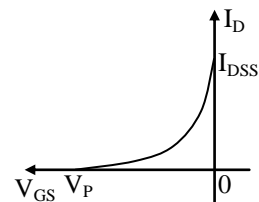
Step(2): Consider the transfer characteristics of a DMOSFET

$$\text{Case (i): } I_D = I_{DSS} \text{ at } V_{GS} = 0 \quad (3)$$

$$\text{Case (ii): } I_D = 0 \text{ at } V_{GS} = V_P \quad (4)$$

$$\Rightarrow V_{GSQ1} = V_P = -4V \quad (5)$$

$$\therefore I_{D1} = 0 \text{ in the ckt given}$$





12. Ans: (b)

Sol: MOD-n ring counter is designed by using 'n' flipflops.

MOD-2n Johnson counter is designed by using 'n' flipflops.

So, MOD-8 ring counter requires 8 flipflops and MOD-8 Johnson counter requires 4 flipflops.

13. Ans: 8.33 (Range: 8 to 8.6)

Sol: According to the concept of virtual ground

$$V_1 = V_2$$

$$\Rightarrow 5V = \frac{30kV_0}{30k + 20k}$$

$$\therefore V_0 = \frac{5V \times 50k}{30k} = 8.33V$$

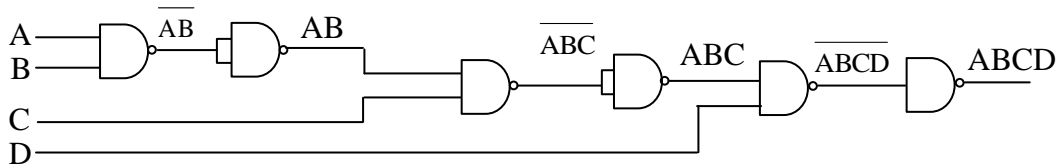
14. Ans: (b)

Sol: Energy $\propto \frac{1}{\text{wave length}}$

$$E = \frac{hc}{\lambda}$$

15. Ans: 6 (No range)

Sol: $Y = ABCD = \overline{\overline{\overline{A} \overline{B} \overline{C} \overline{D}}}$



\therefore 6 NAND gates are required.

16. Ans: (a)

Sol: $|\text{adj}A| = |A|^{n-1}$

$$\Rightarrow 1(12 - 12) - 11(4 - 6) + 3(4 - 6) = |A|^2$$

$$\Rightarrow 22 - 6 = |A|^2$$

$$\therefore |A| = \pm 4$$

17. Ans: 55 (Range: 54.90 to 55)

Sol: Compensator $D(s) = \frac{0.4s + 1}{0.04s + 1} = \frac{1 + aTs}{1 + Ts}$

$$aT = 0.4$$

$$T = 0.04$$

$$\therefore a = 10$$

Maximum phase angle,

$$\phi_m = \sin^{-1} \left(\frac{a-1}{a+1} \right) = 55^\circ$$

18. Ans: (c)

Sol: Given $y(n) - \frac{1}{4}y(n-1) = x(n)$

Apply z transform

$$Y(z) - \frac{1}{4}z^{-1}Y(z) = X(z)$$

$$Y(z) = \frac{X(z)}{1 - \frac{1}{4}z^{-1}}$$

$$x(n) = \delta(n-1)$$

$$X(z) = z^{-1}$$

$$Y(z) = \frac{z^{-1}}{1 - \frac{1}{4}z^{-1}}$$

Apply inverse z transform



$$\left(\frac{1}{4}\right)^n u(n) \leftrightarrow \frac{1}{1 - \frac{1}{4}z^{-1}}$$

From time shifting property

$$\left(\frac{1}{4}\right)^{n-1} u(n-1) \leftrightarrow \frac{z^{-1}}{1 - \frac{1}{4}z^{-1}}$$

$$\text{So, } y(n) = \left(\frac{1}{4}\right)^{n-1} u(n-1)$$

19. Ans: 4 (no range)

$$\text{Sol: } \frac{1}{T_s} = 8000$$

$$\text{So, } T_s = 125 \mu \text{ sec}$$

The separation between samples is $5 \mu \text{sec}$
As the sample is represented by a pulse of $1 \mu \text{sec}$ duration. The separation between two successive pulses is $4 \mu \text{sec}$.

20. Ans: (d)

$$\text{Sol: } (1+t) \frac{dy}{dt} = 4y$$

$$\int \frac{1}{y} dy = \int \frac{4}{1+t} dt$$

$$\log y = 4 \log(1+t) + \log(c)$$

$$y = c(1+t)^4$$

$$y(0) = 1 \Rightarrow 1 = c(1+0)^4 \Rightarrow c = 1$$

$$\Rightarrow y = (1+t)^4$$

21. Ans: (c)

$$\text{Sol: } H_1(z) = \frac{z^2 + 1.5z - 1}{z^2} \text{ and}$$

$$H_2(z) = z^2 + 1.5z - 1$$

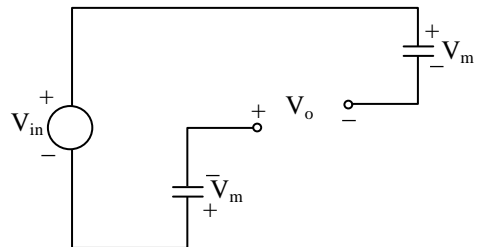
\therefore The zeros of the functions will be identical but not the poles.

22. Ans: (d)

$$\text{Sol: } \frac{E}{R} = \frac{1}{1 - \left[-\frac{4}{s+1} - \frac{4}{s+1} \right]} = \frac{s+1}{s+9}$$

23. Ans: (d)

Sol: During positive cycle D_3 is FB, D_4 is RB. Hence C_1 gets charged to V_m
During Negative cycle D_1 is FB, D_2 is RB. Hence C_2 gets charged to $-V_m$
After the capacitors are charged, the diodes will remain reverse biased



KVL

$$-V_{in} + V_m - V_o - V_m = 0$$

$$V_o = -V_{in}$$

24. Ans: 1 (No range)

Sol: If rank of A is 2, then $|A| = 0$

$$\Rightarrow x^3 - 1$$

$$(x-1)(x^2 + x + 1) = 0$$

$$\Rightarrow x = 1, \frac{-1 \pm \sqrt{3}i}{2}$$

$$\therefore x = 1$$

25. Ans: (d)

Sol: The percent error is

$$e_p = \frac{r - b}{b_{\max} - b_{\min}} \times 100$$

$$= \frac{384 - 379}{440 - 300} \times 100$$

$$= 3.6\%$$

26. Ans: (d)

Sol: Given curve 'C' is a closed curve.

So, we have to evaluate the integral by using Green's theorem.

By Green's theorem, we have



$$\oint_C (M dx + N dy) = \iint_R \left(\frac{\partial N}{\partial x} - \frac{\partial M}{\partial y} \right) dx dy$$

Now, $\oint_C [(x-y)dx + (x+3y)dy]$

$$= \iint_R \left[\frac{\partial}{\partial x} [x+3y] - \frac{\partial}{\partial y} (x-y) \right] dx dy$$

$$= \iint_R [1 - (-1)] dx dy$$

$$= 2 \iint_R 1 dx dy$$

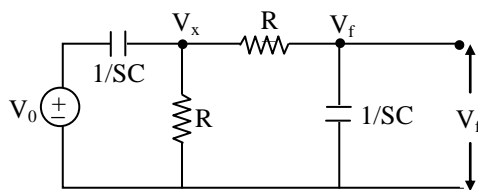
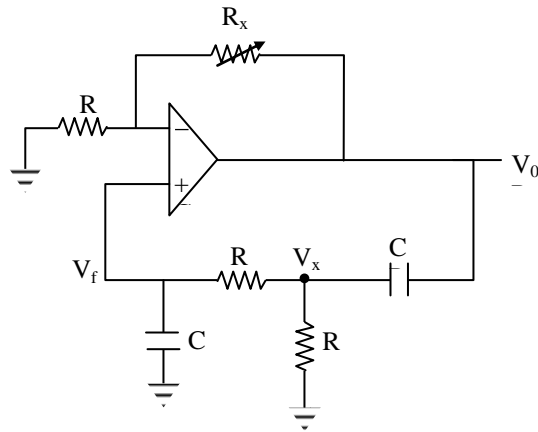
$$= 2(\text{Area of the circle 'C'})$$

$$= 2(\pi r^2)_{r=4}$$

$$= 32\pi$$

27. Ans: (c)

Sol:

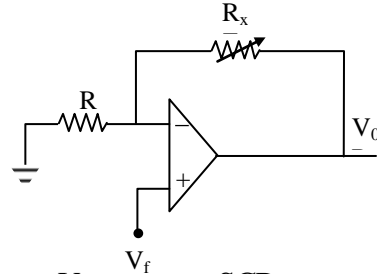


Feedback Network

$$\frac{V_0 - V_x}{\frac{1}{SC}} = \frac{V_x}{R} + \frac{V_x - V_f}{R} \dots\dots (1)$$

and $\frac{V_x - V_f}{R} = \frac{V_f}{\frac{1}{SC}} \dots\dots (2)$

$$\therefore V_x = (1+SRC) V_f \dots\dots(3)$$



$$\Rightarrow \beta = \frac{V_f}{V_0} = \frac{SCR}{S^2C^2R^2 + 3SCR + 1}$$

[\because from equation (1), (2) & (3)]

$$\therefore \beta = \frac{1}{3 + j \left[\omega CR - \frac{1}{\omega CR} \right]}$$

$$\frac{V_0}{V_f} = 1 + \frac{R_x}{R}$$

Since for sustained oscillations $\beta A = 1$

$$\Rightarrow A = \frac{1}{\beta}$$

$$\therefore 1 + \frac{R_x}{R} = 3 + j \left[\omega CR - \frac{1}{\omega CR} \right]$$

Equating img., parts

$$\Rightarrow \omega CR - \frac{1}{\omega CR} = 0$$

$$\Rightarrow f = \frac{1}{2\pi RC} \text{ Hz}$$

$$\& 1 + \frac{R_x}{R} = 3$$

$$\therefore R_x = 2R$$

28. Ans: 560

Sol: Applied pressure $P = 0.8 \times 10^6 \text{ (N/m}^2\text{)}$
Bellows deflection for given pressure

$$\Delta x = 0.125 \text{ mm}$$



$$\text{LVDT sensitivity} \Rightarrow S = 40 \left(\frac{\text{V}}{\text{mm}} \right)$$

$$\text{Sensitivity of the LVDT in } \left(\frac{\text{V}}{\text{N/m}^2} \right) = S \Delta x \frac{1}{P}$$

$$= 40(\text{V/mm}) \times 0.125(\text{mm}) \times \frac{1}{0.8 \times 10^6 (\text{N/m}^2)}$$

$$= 6.25 \times 10^{-6} \frac{\text{V}}{(\text{N/m}^2)}$$

Sensitivity of the LVDT

$$\left(\frac{\mu\text{V}}{\text{N/m}^2} \right) = 6.25 \left(\frac{\mu\text{V}}{\text{N/m}^2} \right)$$

Pressure output of LVDT for

$$3.5\text{V is} = \frac{3.5}{6.25 \times 10^{-6}} = 0.56 \times 10^6 = 560 \text{ KN/m}^2$$

29. Ans: 0.2

Sol: Given that $\frac{dy}{dx} = x^3 - 2y$ ($\because \frac{dy}{dx} = f(x, y)$)

with $y(0) = 0.25$ ($\because y(x_0) = y_0$)

Let $x_0 = 0, y_0 = 0.25$ & $h = 0.1$

Then $x_1 = x_0 + h = 0.1$

The formula for Euler's forward method is

$$y(x_1) \approx y_1 = y_0 + h f(x_0, y_0)$$

$$\Rightarrow y(0.1) \approx y_1 = 0.25 + (0.1) (x_0^3 - 2y_0)$$

$$\Rightarrow y(0.1) \approx y_1 = 0.25 + (0.1) [0 - 2(0.25)]$$

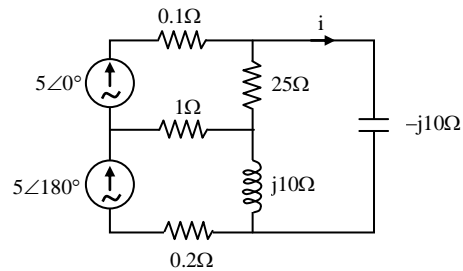
$$\therefore y(0.1) \approx y_1 = 0.25 - (0.1) (0.5)$$

$$= 0.25 - 0.05$$

$$= 0.2$$

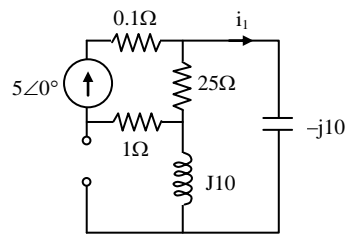
30. Ans: (b)

Sol:

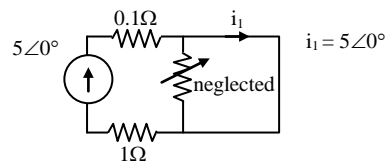


By super position principle $i = i_1 + i_{11}$

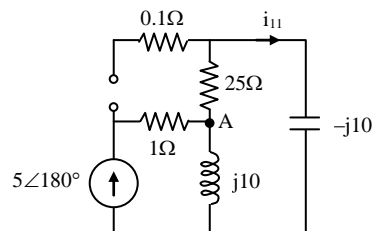
Current i_1 : $5\angle 0^\circ$ source acting alone



$-j10 + j10 = 0 \Omega$ (short circuit)



Current i_{11} : $5\angle 180^\circ$ source acting alone



By current division rule at node A

$$i_{11} = 5\angle 180^\circ \times \frac{j10}{25 + j10 - j10}$$

$$= 2j\angle 180^\circ$$

$$= -2j$$

$$\therefore I = i_1 + i_{11} = 5 - 2j \text{ A}$$



31. Ans: 3750

Sol: Give data:

$$\lambda = 9000 \text{ \AA}$$

$$\Delta = 2\%$$

$$\mu_{\text{core}} = 1.6 = n_1$$

$$\pi = 3$$

$$V = 8$$

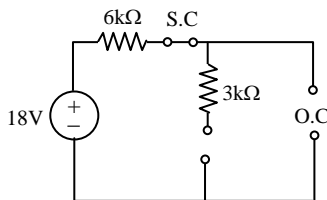
$$V = \frac{2\pi r}{\lambda} n_1 \sqrt{2\Delta}$$

$$8 = \frac{2 \times 3 \times r}{9000 \times 10^{-10}} \times 1.6 \sqrt{2 \times \frac{2}{100}}$$

$$r = 3750 \text{ nm}$$

32. Ans: -60 (No Range)

Sol: At time $t = 0^-$ switch is in open condition



So, L is short circuit, C is open circuit

$$i_L(0^-) = 0$$

$$V_C(0^+) = V_C(0^-) = 18 \text{ V}$$

At $t = 0^+$ switch is closed

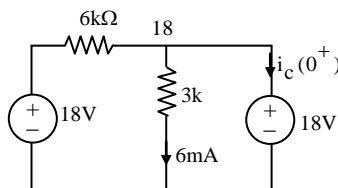
$$I_C(0^+) = C \frac{dv(0^+)}{dt}$$

$$\frac{dv(0^+)}{dt} = \frac{i_C(0^+)}{C}$$

$$i_C(0^+) = -6 \times 10^{-3} \text{ A}$$

$$\frac{dv(0^+)}{dt} = \frac{-6 \times 10^{-3}}{100 \times 10^{-6}}$$

$$\frac{dv(0^+)}{dt} = -60 \text{ V/sec}$$



33. Ans: 16 (Range: 15.9 to 16.2)

Sol: $G(s)H(s) = \frac{10}{(s+2)^4}$

$$\angle G(j\omega)H(j\omega) = \angle \frac{10}{(j\omega+2)^4}$$

$$\text{Gain margin(GM)} = 20 \log \frac{1}{|G(j\omega)H(j\omega)|_{\omega=\omega_{pc}}}$$

$$\angle G(j\omega)H(j\omega)_{\omega=\omega_{pc}} = -180^\circ$$

$$\angle \frac{10}{(j\omega+2)^4} \Big|_{\omega=\omega_{pc}} = -180^\circ$$

$$-4 \tan^{-1} \frac{\omega_{pc}}{2} = -180^\circ$$

$$\Rightarrow \omega_{pc} = 2$$

$$|G(j\omega)H(j\omega)|_{\omega=\omega_{pc}} = \left| \frac{10}{(j(2)+2)^4} \right|$$

$$= \frac{10}{(\sqrt{2^2+2^2})^4} = \frac{10}{64}$$

$$\text{Gain margin} = 20 \log \frac{64}{10}$$

$$= 20[\log 64 - \log 10]$$

$$= 20 [6(0.3) - 1]$$

$$= 16 \text{ (Range: 15.9 to 16.2)}$$

34. Ans: 1.69 (Range: 1.45 to 1.85)

Sol: The value of K can be determined from the following equation:

$$K = \frac{I_{D(\text{on})}}{(V_{GS(\text{on})} - V_{GS(\text{th})})^2}$$

$$= \frac{10 \text{ mA}}{(10 \text{ V} - 1.5 \text{ V})^2} = 1.38 \times 10^{-1} \text{ mA/V}^2$$

$$[\because V_{GS(\text{on})} = 10 \text{ V}]$$

From the circuit, the source voltage is seen to be 0V.

$$\text{Therefore, } V_{GS} = V_G - V_S = V_G - 0 = V_G.$$

The value of $V_G (= V_{GS})$ is given by:



$$V_G \text{ (or } V_{GS}) = \frac{V_{DD}}{R_1 + R_2} \times R_2$$

$$= \frac{10V}{(1+1)M\Omega} \times 1M\Omega = 5V$$

$$I_D = K (V_{GS} - V_{GS(th)})^2$$

$$= (1.38 \times 10^{-1} \text{ mA/V}^2) (5V - 1.5V)^2$$

$$= 1.69 \text{ mA}$$

35. Ans: 0.0045 (Range: 0.004 to 0.005)

Sol: Let X = number of accidents between 5 P.M and 6 P.M.

For Poisson distribution,

$$\lambda = np = (1000) (0.0001) = 0.1$$

$$P(X=x) = \frac{e^{-\lambda} \lambda^x}{x!} \quad (x = 0, 1, 2, \dots)$$

Required Probability = $P(X \geq 2)$

$$= 1 - P(X < 2)$$

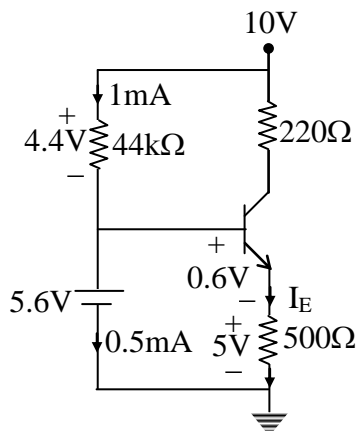
$$= 1 - \{P(X=0) + P(X=1)\}$$

$$= 1 - e^{-0.1} (1 + 0.1)$$

$$= 0.0045$$

36. Ans: 0.95 (No range)

Sol: Zener diode is in breakdown, replace it with a voltage source of value $V_Z = 5.6V$ & $V_{BE} = 0.6V$



Applying KCL at Base, we get

$$-1 \text{ mA} + 0.5 \text{ mA} + I_B = 0$$

$$I_B = 0.5 \text{ mA}$$

$$I_E = \frac{5}{500\Omega} = 10 \text{ mA}$$

$$I_E = (\beta + 1) I_B$$

$$\beta + 1 = \frac{I_E}{I_B}$$

$$\beta + 1 = \frac{10}{0.5}$$

$$\beta + 1 = 20$$

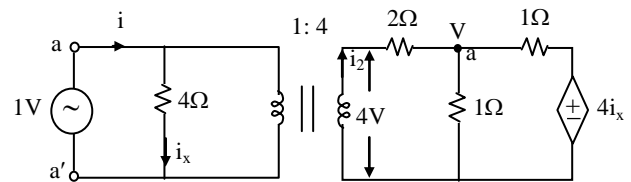
$$\beta = 19$$

$$\alpha = \frac{\beta}{1 + \beta} = \frac{19}{20} = 0.95$$

37. Ans: 171 (Range: 165 to 175)

Sol: Now 1V is applied at primary side a-a'

$$R = \frac{1}{i}$$



By transformation ratio

$$K = \frac{4}{1} = 4$$

$$\frac{V_2}{V_1} = k = 4$$

$$V_2 = 4V$$

$$i_x = \frac{1}{4} \text{ A}$$

Apply KCL at node-a,

$$\frac{V-4}{2} + \frac{V}{1} + \frac{V-4i_x}{1} = 0$$

$$V - 4 + 2V + 2V - 8i_x = 0$$

$$5V = 4 + 8i_x = 4 + 8 \left(\frac{1}{4}\right) = 6$$

$$V = \frac{6}{5} \text{ V}$$



i_2 is current flows through secondary winding then

$$i_2 = \frac{4 - \frac{6}{5}}{2} = \frac{14}{10} = 1.4 \text{ A}$$

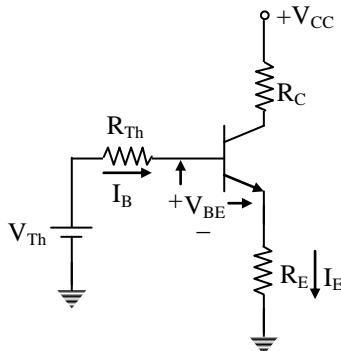
This i_2 transferred to primary then $i'_2 = 1.4 \times 4$

$$\therefore i = i_x + i'_2 = \frac{1}{4} + 1.4 \times 4 = 5.85 \text{ A}$$

$$\therefore R = \frac{1}{i} = \frac{1}{5.85} = 171 \text{ m}\Omega$$

38. 4.92 (Range: 4.8 to 5.1)

Sol:



fig(a): Thevenin equivalent of the given circuit

Step (1):

KVL for the input loop of circuit shown in fig (a)

$$V_{Th} - I_B R_{Th} - V_{BE} - I_E R_E = 0 \text{ ----- (1)}$$

$$I_E = \frac{V_{Th} - V_{BE}}{R_E + \frac{R_{Th}}{1 + \beta}} \text{ ---- (2) [}\because I_B = \frac{I_E}{1 + \beta}\text{]}$$

$$V_{Th} = \frac{V_{CC} R_2}{R_1 + R_2}$$

$$R_{Th} = R_1 // R_2$$

Step (2):

$$\text{But } I_E R_E = \frac{V_{CC}}{3} = \left[\frac{V_{Th} - V_{BE}}{R_E + \frac{R_{Th}}{1 + \beta}} \right] R_E \text{ ---- (3)}$$

(\because Given that voltage drop across

$$R_E = \frac{V_{CC}}{3}$$

$$\left\{ \frac{(V_{Th} - V_{BE})}{R_E \left[1 + \left(\frac{R_{Th}}{R_E} \right) \left(\frac{1}{1 + \beta} \right) \right]} \right\} R_E = \frac{V_{CC}}{3} \text{ ---- (4)}$$

$$V_{Th} = \frac{V_{CC}}{3} \left[1 + \left(5.73 \times \frac{1}{101} \right) \right] + V_{BE} \text{ ---- (5)}$$

$$V_{Th} = 4.92693 \text{ V ---- (6)}$$

39. Ans: (b)

Sol: Given $v = y + e^{-x} \cos y$

$$\Rightarrow v_x = -e^{-x} \cos(y) \text{ and } v_y = 1 - e^{-x} \sin(y)$$

Consider $du = (u_x) dx + (u_y) dy$

$$= (v_y) dx + (-v_x) dy$$

$$\Rightarrow du = (1 - e^{-x} \sin y) dx + (e^{-x} \cos y) dy$$

$$\Rightarrow \int du = \int (1 - e^{-x} \sin y) dx + \int 0 dy + k$$

$$\Rightarrow u = x + e^{-x} \sin y + k$$

Now the required analytic function $f(z)$ is given by $f(z) = u + iv$

$$\Rightarrow f(z) = (x + e^{-x} \sin y + k)$$

$$+ i (y + e^{-x} \cos y)$$

$$\therefore f(z) = z + i e^{-z} + k$$

40. Ans: (b)

Sol: For break point,

$$\frac{dk}{ds} = 0$$

$$\frac{d}{ds} \left(\frac{1}{G(s)H(s)} \right) = 0$$

$$\frac{d}{ds} (s(s+6)(s^2+4s+13)) = 0$$

$$(s^2+6s)[2s+4] + (s^2+4s+13)[2s+6] = 0$$

$$2s^3 + 16s^2 + 24s + 2s^3 + 6s^2 + 8s^2 + 24s + 26s + 78 = 0$$

$$4s^3 + 30s^2 + 74s + 78 = 0$$

$$f(s) = 2s^3 + 15s^2 + 37s + 39 = 0 \text{ ---- (1)}$$

$$f(-5) = -21$$



$$f(-4) = 3$$

$$f(-3) = 9$$

As there is a sign change in between $-5, -4$, one root is on real axis, which is in between

$-5, -4$. Three real axis break points is not possible.

41. Ans: (b)

Sol: $f(t) = \left(-\frac{t}{T} + 1\right) [u(t) - u(t - T)]$

$$f(t) = \left(-\frac{t}{T}\right) [u(t) - u(t - T)] + [u(t) - u(t - T)]$$

$$[u(t) - u(t - T)] \leftrightarrow \frac{1}{s} - \frac{e^{-sT}}{s}$$

From differentiation in s-domain property

$$\left(-\frac{t}{T}\right) [u(t) - u(t - T)] \leftrightarrow \frac{1}{T} \frac{d}{ds} \left(\frac{1}{s} - \frac{e^{-sT}}{s}\right) = \frac{1}{T} \left[\frac{-1}{s^2} - \frac{(se^{-sT}(-T) - e^{-sT})}{s^2}\right]$$

$$\left(-\frac{t}{T}\right) [u(t) - u(t - T)] \leftrightarrow \frac{1}{T} \left[\frac{-1}{s^2} + T \frac{e^{-sT}}{s} + \frac{e^{-sT}}{s^2}\right] = \frac{e^{-sT}}{s} + \frac{e^{-sT}}{Ts^2} - \frac{1}{s^2T}$$

$$F(s) = \frac{1}{s} - \frac{e^{-sT}}{s} + \frac{e^{-sT}}{s} + \frac{e^{-sT}}{Ts^2} - \frac{1}{s^2T} = \frac{1}{s} + \frac{e^{-sT}}{Ts^2} - \frac{1}{s^2T} = \frac{1}{s^2T} [sT - 1 + e^{-sT}]$$

42. Ans: 3 (No range)

Sol: $J_0 = \bar{Q}_1; K_0 = 1; J_1 = Q_0; K_1 = 1$

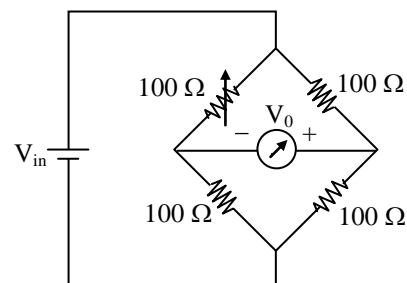
J_0	K_0	J_1	K_1	Q_0	Q_1
				0	0
1	1	0	1	1	0
1	1	1	1	0	1
0	1	0	1	0	0
1	1	0	1	1	0

→ It is a mod-3 counter.

43. Ans: (c)

Sol: $V_{in} = 15 \times 10^{-3} [100 + 100]$

$V_{in} = 3 \text{ Volts}$



% change in "R_s" is 5% ⇒ $100 \times \frac{5}{100} = 5 \Omega$

$R_s = 105 \Omega$

$$V_0 = \frac{3 \times 100}{200} - \frac{3 \times 100}{205} = 300 \left[\frac{1}{200} - \frac{1}{205} \right]$$

$= 0.03658 = 36.58 \text{ mV}$



44. Ans: (c)

Sol: By giving different sets of input values and Q(t) (present state) we have to determine next state Q(t+1)

X	Y	Q(t)	S	R	Q(t+1)
0	0	0	1	0	1
0	0	1	1	0	1
0	1	0	0	1	0
0	1	1	0	1	0
1	0	0	0	0	0
1	0	1	0	0	1
1	1	0	1	1	×
1	1	1	1	1	×

K-map for Q (t + 1)

		YQ(t)			
		00	01	11	10
X	0	1	1		
	1		1	×	×

$$\therefore Q(t+1) = \bar{X}\bar{Y} + \bar{Y}Q(t) \quad \text{or}$$

$$Q(t+1) = \bar{X}\bar{Y} + XQ(t)$$

45. Ans: 0.5

Sol: Given y(n) be a 4 point circular convolution of x(n) & h(n)

$$y(n) = x(n) \text{ circular convolution } h(n)$$

Apply DFT

$$Y(k) = X(k) H(k)$$

$$Y(k) = \{1, -2, 1, -2\} \cdot \{1, j, 1, -j\}$$

$$= \{1, -2j, 1, 2j\}$$

$$\text{IDFT } y(n) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k) e^{j\frac{2\pi}{N}nk}$$

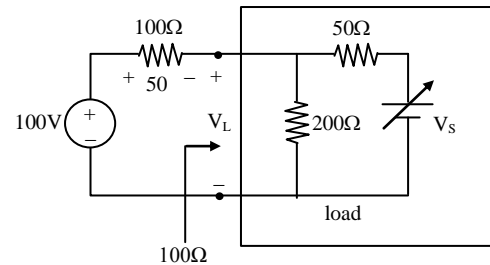
$$y(0) = \frac{1}{N} \sum_{k=0}^{N-1} Y(k)$$

$$N = 4$$

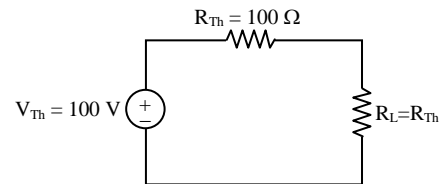
$$y(0) = \frac{1}{4} \sum_{k=0}^3 Y(k) = \frac{1}{4} [1 - 2j + 1 + 2j] = \frac{2}{4} = \frac{1}{2}$$

46. Ans: 37.5 (Range: 37 to 38)

Sol: Applying thevenin's theorem in the source side, the circuit can be redrawn as



We know that for maximum power transform, the load resistance is equal to Thevenin's resistance.



$$\text{Voltage drop across load, } V_L = \frac{V_{Th} \times R_L}{R_{Th} + R_L}$$

$$\Rightarrow V_L = \frac{V_{Th} \times R_L}{2R_{Th}}$$

$$V_L = \frac{V_{Th}}{2}$$

Consider above figure, for maximum power transferred to load

$$V_L = \frac{V_{Th}}{2} = \frac{100}{2} = 50 \text{ Volts}$$

Applying nodal analysis, we get

$$\Rightarrow \frac{50}{100} = \frac{50}{200} + \frac{50 - V_s}{50}$$



$$\Rightarrow \frac{50 - V_s}{50} = \frac{1}{4}$$

$$\Rightarrow 50 - V_s = 12.5$$

$$\Rightarrow V_s = 37.5 \text{ V}$$

47. Ans: (c)

Sol: $A_c \cos 2\pi f_c t + \frac{A_c \mu}{2} \cos 2\pi(f_c + f_m)t + \frac{A_c \mu}{2} \cos 2\pi(f_c - f_m)t$

USB is attenuated by a factor of '2'

$$= A_c \cos 2\pi f_c t + \frac{A_c \mu}{4} \cos 2\pi(f_c + f_m)t + \frac{A_c \mu}{2} \cos 2\pi(f_c - f_m)t$$

$$= \cos 2\pi f_c t + \frac{1}{8} \cos 2\pi(f_c + f_m)t + \frac{1}{4} \cos 2\pi(f_c - f_m)t$$

The inphase component is

$$\left[1 + \frac{1}{8} \cos 2\pi f_m t + \frac{1}{4} \cos 2\pi f_m t \right] \cdot \cos 2\pi f_c t = \left[1 + \frac{3}{8} \cos 2\pi f_m t \right]$$

48. Ans: (b)

Sol: • (SP) = 8086H

• (DE) = 8085H

• (HL) exchanged with (DE)

After execution

(HL) = 8085H, (DE) = $\times \times \times \times$ H

• (HL) = 8085H = 1000 0000 1000 0101

(SP) = 8086H = 1000 0000 1000 0110

(HL) = 010BH = 0000 0001 0000 1011

• (HL) = 010BH copied into (SP)

\Rightarrow (SP) = 010BH

• 8085 microprocessor calls delay subroutine and after execution of subroutine, microprocessor returns to main program.

SP contents decremented by 2 for CALL operation

SP contents incremented by 2 for RETURN operation

i.e., (SP) = 010BH - 2 + 2

= 010BH

• 8085 microprocessor pushes DE pair contents to stack. SP contents will be decremented by 2 for PUSH operation.

(SP) = 010BH - 2 = 0109H

• 8085 microprocessor executes RST 7 software interrupt where it calls respective ISR and returns to main execution.

(SP) = 0109H - 2 + 2 = 0109H

• 8085 microprocessor contents of top 2 stack locations back into PSW. SP contents will be incremented by 2 for POP operation.

(SP) = 0109H + 2 = 010BH

• (SP) = 010BH

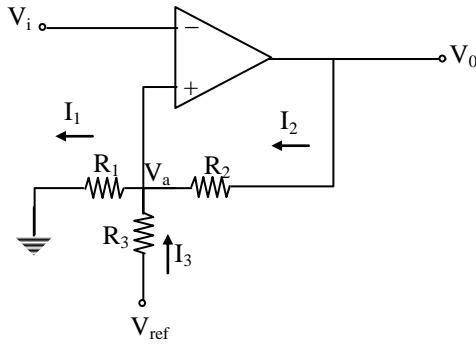


49. Ans: 2.19 (2.18 to 2.20)

Sol: Apply KCL at V_a

$$I_1 = I_2 + I_3$$

$$\frac{V_a}{R_1} = \frac{V_{ref} - V_a}{R_3} + \frac{V_0 - V_a}{R_2}$$



$$V_a \left[\frac{1}{R_1} + \frac{1}{R_2} + \frac{1}{R_3} \right] = \frac{V_{ref}}{R_3} + \frac{V_0}{R_2}$$

$$V_a = V_{UTP} \text{ when } V_0 = +V_{sat}$$

$$V_a = V_{LTP} \text{ when } V_0 = -V_{sat}$$

$$V_a = \frac{\frac{V_{ref}}{R_3} + \frac{V_{sat}}{R_2}}{\left[R_1^{-1} + R_2^{-1} + R_3^{-1} \right]}$$

$$V_{UTP} = \frac{\frac{-10}{10} + \frac{12}{50}}{\left[\frac{1}{10} + \frac{1}{10} + \frac{1}{50} \right]} = -3.4545 \text{ V}$$

$$V_{LTP} = -5.6363 \text{ V}$$

$$V_H = V_{UTP} - V_{LTP}$$

$$V_H \approx 2.19$$

50. Ans: (d)

Sol: $X(\omega) = \frac{1}{(2 + j\omega)^2}$

$$H(\omega) = \frac{1}{4 + j\omega}$$

$$Y(\omega) = X(\omega) \cdot H(\omega) = \frac{1}{(4 + j\omega)(2 + j\omega)^2}$$

$$= \frac{1/4}{4 + j\omega} - \frac{1/4}{2 + j\omega} + \frac{1/2}{(2 + j\omega)^2}$$

$$y(t) = \frac{1}{4}e^{-4t}u(t) - \frac{1}{4}e^{-2t}u(t) + \frac{1}{2}te^{-2t}u(t)$$

$$y(3) = \frac{1}{4}e^{-12} - \frac{1}{4}e^{-6} + \frac{3}{2}e^{-6} = \frac{1}{4}e^{-12} + \frac{5}{4}e^{-6}$$

51. Ans: (c)

Sol: $P_1 = 600 \text{ W}$

$P_2 = 300 \text{ W}$

$$\therefore \theta = \tan^{-1} \left(\sqrt{3} \frac{P_1 - P_2}{P_1 + P_2} \right)$$

$$= \tan^{-1} \left[\sqrt{3} \left(\frac{600 - 300}{600 + 300} \right) \right] = 30^\circ$$

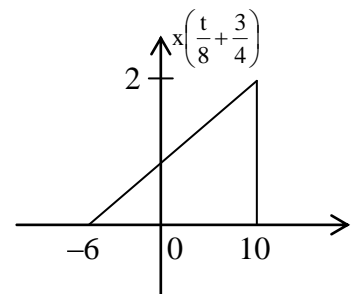
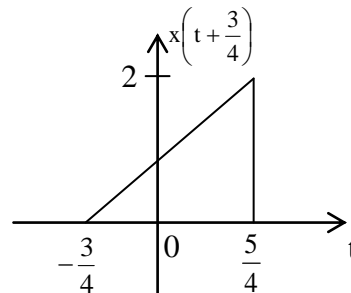
$$P = P_1 + P_2 = 600 + 300 = 900 \text{ W}$$

$$\tan \theta = \frac{Q}{P} \Rightarrow \tan 30^\circ = \frac{Q}{900}$$

$$\Rightarrow Q = \frac{900}{\sqrt{3}} = 300\sqrt{3} \text{ VAR}$$

52. Ans: (d)

Sol:



53. Ans: (a)

Sol: Given $(2xy - 9x^2)dx + (2y + x^2 + 1)dy = 0$

Here, $M = 2xy - 9x^2$ and $N = 2y + x^2 + 1$



Now, $\frac{\partial M}{\partial y} = 2x = \frac{\partial N}{\partial x}$

∴ The given D.E is exact

Now the general solution of the given D.E is

$$\int(2xy - 9x^2)dx + \int(2y + 0 + 1)dy = C$$

$$\Rightarrow x^2y - 9\frac{x^3}{3} + y^2 + y = C \dots\dots (1)$$

but $y = -3$ at $x = 0$

Now (1) becomes

$$0 - 0 + 9 - 3 = C$$

$$\Rightarrow C = 6$$

∴ The solution of a given D.E is

$$x^2y - 3x^3 + y^2 + y = 6$$

54. Ans: 0.02

Sol:

$$d = 5 \times 10^{-2} \text{ m,}$$

$$B = 0.1 \text{ T,}$$

$$E = 0.1 \times 10^{-3} \text{ V}$$

$$v = \frac{E}{Bl} = \frac{0.1 \times 10^{-3}}{(0.1)(5 \times 10^{-2})} = 0.02 \text{ m/s}$$

55. Ans:0 (No range)

Sol: $\delta(t^2 - a^2) = \frac{1}{2|a|} [\delta(t+a) + \delta(t-a)]$

$$I = \int_{-1}^1 \frac{1}{4} [\delta(t-2) + \delta(t+2)] dt$$

$$I = \frac{1}{4} \int_{-1}^1 \delta(t-2) dt + \frac{1}{4} \int_{-1}^1 \delta(t+2) dt$$

From sifting property

$$\int_{t_1}^{t_2} x(t)\delta(t-t_0)dt = x(t_0) \quad t_1 \leq t_0 \leq t_2$$

$$= 0 \quad \text{otherwise}$$

$$I = 0 + 0 = 0$$

56. Ans: (d)

Sol: (PART AND THE WHOLE) A fragment is a piece of broken bone; a shard is a piece of broken pottery. (d)

57. Ans: (a)

58. Ans: (d)

Sol: irretrievably means impossible to recover or get back, so irrevocably is the correct synonym, which means not capable of being changed : impossible to revoke.

59. Ans: (b)

Sol: Indiscriminate (adj.) means not discriminating or choosing randomly; haphazard; without distinction.

60. Ans: (a)

Sol: $a_0 = 1$; $a_n = 2a_{n-1}$ if n is odd

$$a_n = a_{n-1} \text{ if n is even}$$

$$a_{100} = a_{100-1} = a_{99} = 2.a_{99-1}$$

$$= 2.a_{99} = 2.a_{98-1} = 2a_{97}$$

$$= 2.2a_{97-1} = 2^2.a_{96} \dots\dots\dots 2_{50}.a_0 = 2^{50}$$

61. Ans: (c)

Sol: $A = 1$; $B = 1$

$$(a) B = B + 1 = 2$$

$$(b) \& (c) A = A \times B = 1 \times 2 = 2$$

$$\text{Step 2: } B = 2 + 1 = 3; A = A \times B = 2 \times 3 = 6$$

$$\text{Step 3: } B = 3 + 1 = 4; A = A \times B = 6 \times 4 = 24$$

$$\text{Step 4: } B = 4 + 1 = 5; A = 24 \times 5 = 120$$

$$\text{Step 5: } B = 5 + 1 = 6; A = 120 \times 6 = 720$$

62. Ans: (a)

Sol: Ratio of efficiency (P & Q) = 2 : 1

Ratio of efficiency (P + Q, R) = 3 : 1

If R does 1 unit work, then P& Q together do 3 units.



Out of 3 units, P does 2 units and Q does 1 unit.

\therefore Ratio of efficiency (P, Q & R) = 2 : 1 : 1
Hence, earnings should be divided in the ratio is 2 : 1 : 1

63. Ans: (c)

Sol: In 1972, A was as old as the number formed by the last two digits of his year of birth.

So, A was born in 1936 (as in 1972, he is 36 yrs older also, last two digits of 1936 are 36).

Hence, B was born in $1936 + 15 = 1951$
so, he is 21 yrs old in 1972

64. Ans: (b)

Sol: Difference (in thousands) between the numbers of customers in the 2 complexes in:

January: $22 - 20 = 2$

February: $25 - 24 = 1$

March: $20 - 15 = 5$

April: $28 - 25 = 3$

May: $20 - 14 = 6$ [Max]

June: $20 - 15 = 5$

65. Ans: (b)

Sol: The issue is more about punishing criminals, and so punishment is more important than crime prevention (correct answer B).