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# **ACE Pre-GATE 2017**

### Q.1 – Q.5 Carry One Mark Each

**Branch: INST** 

01.	01. Reaching a place of appointment on Friday. I found that I was two days earlier than the sc									
	day. If I had reached on the following Wednesday then how many days late would I have been?									
	(a) one (b) two (c) thr	ee (d) four								
01.	01. Ans: (c)	0.1								
Sol:	Sol: Friday → 2 days earlier	3								
	Therefore, scheduled day = Friday $+ 2$ = Sunday									
	Sunday + 3 = Wednesday									
	Therefore, I would have been late by 3 days									
02.	11 1									
	sentence.	5								
	The bus stopped to more passengers.	The bus stopped to more passengers.								
	(a) Take in (b) Take on (c) Take	ke up (d) Take for								
02.	02. Ans: (b)									
0.2	O2 Change the appropriate contains from the fellowing	unti anno								
03.		pptions.								
	<ul><li>(a) She has been discharged since</li><li>(b) She has since been discharged</li></ul>									
	(c) She has been since discharged									
	(d) She since has been discharged									
03.	03. Ans: (b)									

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04.	Fill in the blank with an appropriate phrase.								
	The jet into the air	r.							
	(a) Soared	(b) Soured							
	(c) Sourced	(d) Sored							
04.	Ans: (a)								
05.	. Choose the most appropriate word from the options given below to complete the following								
	sentence.								
	If I had known that you were	coming, I	you at the airport.						
	(a) Would meet	(b) Would have met							
	(c) Will have met	(d) Had met	*C <sub>4</sub>						
05.	Ans: (b)	· 1	THE STATE OF THE S						

# **SHORT TERM BATCHES FOR GATE+PSUs - 2018**

HYDE	RABAD
29 <sup>TH</sup> APRIL 2017	06 <sup>™</sup> MAY 2017
13 <sup>TH</sup> MAY 2017	18 <sup>TH</sup> MAY 2017
O1 <sup>st</sup> JU	NE 2017

NEW BATCHES FOR ESE | GATE | PSUs - 2018 CONTACT : 040 - 23234418 / 19 / 20 www.aceenggacademy.com



# Q.6 - Q.10 Carry two marks each

Which of the following can be logically inferred from the given statement.

"No other studied medicine except Helen"

- (a) Helen only studied medicine
- (b) Only Helen studied medicine
- (c) Helen studied only medicine
- (d) Helen studied medicine only
- 06. Ans: (b)

The average electricity bill of a household for January to June is ₹ 980, for July to September is 07. ₹ 670, for October to December is ₹ 720. If the family goes on vacation for June and July and no electricity is used, what would be the average electricity bill for that year?

(a) ₹ 500

(b) ₹ 600

(c) ₹ 700

(d) ₹ 800

07. Ans: (c)

Sol: Average electricity bill from January to June = ₹ 980

∴ Total electricity bill from January to May =  $980 \times 5 = ₹4900$ 

(As no electricity is used in June)

Similarly, total electricity bill from August to September (as no electricity is used in July)

And total electricity bill from October to December =  $720 \times 3 = 2160$ 

Therefore, total electricity bill from January to December = 4900 + 1340 + 2160 = 8400

Thus, average electricity bill for the whole year =  $\frac{8400}{12}$  = ₹700

08. The following question has four statements of three segments each. Choose the alternative where the third segment in the statement can be deduced using both the preceding two but not just from one of them.

A. Sonia is an actress. Some actresses are pretty. Sonia is pretty.

B. All actors are pretty. Manoj is not an actor. Manoj is not pretty



- C. Some men are cops. Some men are brave. Some brave people are cops.
- D. All cops are brave. Some men are cops. Some men are brave.
- (a) only C
- (b) only A
- (c) only D
- (d) B and C

08. Ans: (c)

**Sol:** Statements:

All cops are brave

Some men are cops



Conclusion:

Some men are brave (True)

Hence, only D follows.

- 09. A contractor, who got the contract for building the flyover, failed to construct the flyover in the specified time and was supposed to pay ₹ 50,000 for the first day of extra time. This amount increased by ₹ 4,000 each day. If he completes the flyover after one month of stipulated time, he suffers a loss of 10% in the business. What is the amount he received for making the flyover in crores of rupee? (One month = 30 days)
  - (a) 3.1

- (b) 3.24
- (c) 3.46
- (d) 3.68

09. Ans: (b)

**Sol:** The sum of money that the contractor was supposed to pay for the period of an month over the stipulated time is

$$= S_n = \frac{n}{2} [2a + (n-1)d]$$

$$a = 50,000, n = 30, d = 4000$$



$$S_{30} = \frac{30}{2} [2 \times 50,000 + (30 - 1) \times 4000] = 15[100,000 + 29 \times 4000]$$

₹ 
$$3240000 = ₹ 32.4$$
 lakhs

Loss in the business = 10%

- ∴ Amount he received for making the flyover =  $\frac{3240000}{0.1}$  = 32400,000 = ₹ 3.24 crores
- 10. Study the following pie chart and table carefully to answer the following question that follows: Percentage break up of employees working in various departments of an organisation and the ratio of men to women in them.

Total number of employees = 1800

# Percentage break up of employees:



#### Ratio of men to women

Department	Men	Women
Production	11	1
HR	1	3
IT	5	4
Marketing	7	5
Accounts	2	7

What is the number of men working in the marketing department?

(a) 132

(b) 174

- (c) 126
- (d) 189

10. Ans: (d)

**Sol:** Number of men working in the marketing department =  $1800 \times \frac{18}{100} \times \frac{7}{12} = 189$ 



# Q.11 – Q.35 Carry one mark each.

The signal  $x(t) = \cos(50\pi t) + \cos(80\pi t)$  is sampled at 200Hz. The minimum number of samples required to prevent leakage is \_\_\_\_\_

**Ans: 40** 

**Sol:** 
$$\frac{\omega_1}{2\pi} = \frac{50\pi/200}{2\pi} = \frac{1}{8} = \frac{5}{40}$$

$$\frac{\omega_2}{2\pi} = \frac{80\pi/200}{2\pi} = \frac{1}{5} = \frac{8}{40}$$

- :. The minimum number of samples required to prevent leakage is 40.
- An n-channel JFET has a gate cut-off voltage of -10V. If -1V is applied at the gate terminal the minimum drain to source voltage required to obtain maximum drain resistance is \_\_\_\_\_(V) (Neglect channel length modulation)

Ans: 9

**Sol:** Drain resistance will be maximum at saturation

Minimum voltage ( $V_{DS}$ ) for saturation =  $V_{GS} - V_P$ 

Given  $V_P$  = Gate cut-off voltage = -10 V

$$V_{DS} = V_{GS} - V_P = -1 - (-10) = +9 \text{ V}$$

Temperature is measured by a sensor with an output of 0.01 V/°C. The required ADC word size to measure 0 °C to 100 °C with 0.01 °C resolution is

**Ans: 14** 

**Sol:** Number of steps of 0.01 °C in 100 °C =  $\frac{100^{\circ} \text{C}}{0.01^{\circ} \text{C}}$  = 10000 steps

$$2^n = 10000$$

n = word size

 $n \cong 14$  bits



Multimode step index fiber of step index 4.5% with core radius  $100/\pi$  µm & wavelength of 1 µm. If core refractive index is 1.6 then V-number is

**Ans: 96** 

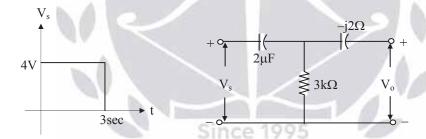
Sol: 
$$V = \frac{2\pi r}{\lambda} n_1 \sqrt{2\Delta} = \frac{2\pi \times \frac{100}{\pi} \times 10^{-6} \times 1.6\sqrt{2 \times 0.045}}{1 \times 10^{-6}} = 96$$

In 3-φ & 4 wire balanced system, the maximum number of wattmeters required to calculate the power is

Ans: 3

Sol: For 3-φ & 4 wire system, as neutral is present, we can use maximum 3 wattmeters

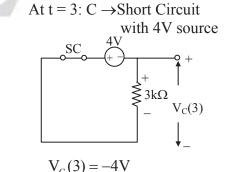
A square pulse of 4V amplitude is applied to RC circuit shown in figure. The capacitor is initially 16. uncharged. The output voltage  $V_0$  at time  $t = 3 \sec is$  \_\_\_\_ (in Volts).

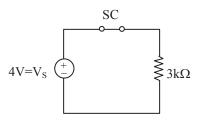


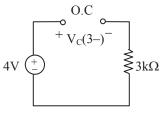
Ans: -4V

Sol:

At t = 0: C  $\rightarrow$ Short Circuit At t = (3-): C  $\rightarrow$ Open Circuit





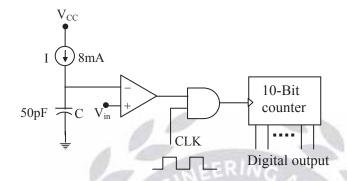


 $V_C(3-) = V_C(3+) = 4V$ 

Note: In the circuit  $-j2\Omega$  capacitor Voltage is zero for all time since there is no current is flowing through that capacitor. Hence it acts as a Short Circuit.



17. A single slope ADC shown below is used to convert an analog input of 16V to digital. Determine the digital output (in decimal) if the clock frequency is 500 MHz. Initial voltage across capacitor is 0V



**Ans: 50** 

Sol:

• Initially  $V_c = 0V$ ;  $V_{in} = 16V$ 

⇒ comparator output = 1 & clock pulses reach the Counter

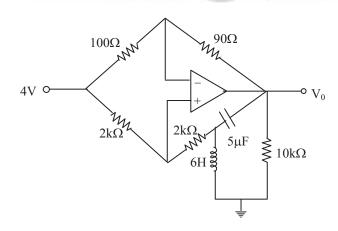
• Time required for the capacitor to charge to 16V is It = CV

$$\Rightarrow t = \frac{CV}{I} = \frac{50 \times 10^{-12} \times 16}{8 \times 10^{-3}} = 100 \text{ns}$$

Clock period T = 
$$\frac{1}{500 \times 10^6}$$
 = 2 ns

• Digital output = 
$$\frac{100 \text{ ns}}{T} = \frac{100 \text{ ns}}{2 \text{ ns}} = (50)_{10}$$

18. If the Op-amp is ideal. Find the current (in Amp) across  $90\Omega$  resistor.





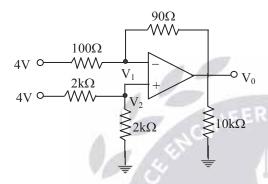
Ans: 0.02 Range: 0.01 to 0.03

Sol: Since, DC 4V is given as input, no switch (transient) is provided

 $\Rightarrow$  we can proceed under steady state.

 $C \rightarrow$  behaves as open &  $L \rightarrow$  Acts as short

the circuit is  $\rightarrow$ 



Hence the circuit behaves as a subtractor.

:. From superposition,

$$V_0 = \frac{-90}{100} [4] + \left[ 1 + \frac{90}{100} \right] \left[ \frac{4[2k]}{2k + 2k} \right] \text{Volts}$$
  
= -3.6 + 3.8  
= 0.2 V

Hence, from circuit, as 
$$V_2 = \frac{4(2k)}{2k + 2k} = V_1$$

$$\therefore V_1 = 2V$$

$$\therefore I_{90\Omega} = \frac{V_1 - V_0}{90} = 0.02A$$

- If A, B & C are  $n \times n$  matrices and |A| = 2, |B| = 3 & |C| = 5 then the value of  $|A^2BC^{-1}| = ?$
- Ans: 2.4 19. Range: 2 to 3

**Sol:** 
$$|A^2 B C^{-1}| = \frac{|A||A||B|}{|C|} = \frac{2 \times 2 \times 3}{5} = \frac{12}{5} = 2.4$$



20. What is the value of  $\underset{x\to\infty}{\text{Lt}} \sqrt{(x^2+x+1)} - x$ 

Sol: 
$$\underset{x \to \infty}{\text{Lt}} \sqrt{(x^2 + x + 1)} - x = \underset{x \to \infty}{\text{Lt}} \left[ \sqrt{x^2 + x + 1} - x \right] \left[ \frac{\sqrt{x^2 + x + 1} + x}{\sqrt{x^2 + x + 1} + x} \right]$$
$$= \frac{1 + 0}{\sqrt{1 + 0 + 0} + 1} = \frac{1}{2} = 0.5$$

21. An Engineer applies the input  $r(t) = 2\sin(t - 1.55)$  to a chemical process and measures—the output as  $y(t) = 0.4\sin(t - 1.55)$ . What is the gain of the system?

Ans: 0.2

**Sol:** Input 
$$r(t) = 2\sin(t-1.55)$$

Output 
$$y(t) = 0.4\sin(t-1.55)$$

For sinusoidal input, the output is

$$y(t) = A \times M \sin(t \pm \theta \pm \phi)$$

$$A \times M = 0.4$$
 (A  $\rightarrow$  Amplitude of input), A = 2

$$2M = 0.4 \Rightarrow M = 0.2$$

22. Find the difference equation relating input x(n) and output y(n) for the realization shown in figure. The filter impulse response is  $h(n) = 0.5\delta(n) + 0.5\delta(n-1)$ .

$$x(n)$$
 $+$ 
 $\sum$ 
filter
 $z^{-1}$ 

Option A: 
$$y(n) + y(n-1) + y(n-2) = x(n) + x(n-1)$$

Option B: 
$$y(n) + 0.5y(n-1) + 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

Option C: 
$$y(n) - 0.5y(n-1) - 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

Option D: 
$$1.5y(n) + 0.5y(n-1) = 0.5x(n) + 0.5x(n-1)$$



# Ans: (B)

Sol:

Assume filter transfer function is  $H_1(z)$ .

$$H_1(z) = 0.5 + 0.5z^{-1}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{H_1(z)}{1 + z^{-1}H_1(z)} = \frac{0.5 + 0.5z^{-1}}{1 + z^{-1}(0.5 + 0.5z^{-1})}$$

$$\frac{Y(z)}{X(z)} = \frac{0.5 + 0.5z^{-1}}{1 + 0.5z^{-1} + 0.5z^{-2}}$$

$$Y(z)[1+0.5z^{-1}+0.5z^{-2}] = 0.5X(z) + 0.5z^{-1}X(z)$$

$$y(n) + 0.5y(n-1) + 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

# **Distractor Logic:**

Option A: If we miss 0.5 in filter impulse response.

Option B: Correct option

Option C: If we feel positive feedback system, wrong interpretation of taking negative coefficients.

Option D: In the feedback multiplier of  $h_1(n)$  if we miss  $z^{-1}$ 

An RC low pass filter has the impulse response  $h(t) = e^{-t} u(t)$ . The response of the system due 23. to the input  $x(t) = e^{2t} u(-t)$  is\_\_\_\_\_

Option A: 
$$\frac{1}{3}e^{2t}u(-t) + \frac{1}{3}e^{-t}u(t)$$

Option B: 
$$\frac{-1}{3}e^{2t}u(t) - \frac{1}{3}e^{-t}u(-t)$$

Option C: 
$$-\frac{1}{3}e^{2t}u(-t) - \frac{1}{3}e^{-t}u(t)$$

Option D: 
$$e^{2t}u(-t) + e^{-t}u(t)$$

Ans: (A)

**Sol:** 
$$H(s) = \frac{1}{s+1}$$
;  $\sigma > -1$ 

$$X(s) = \frac{-1}{s-2}$$
;  $\sigma < 2$ 



Output ROC = 
$$(\sigma > -1) \cap (\sigma < 2) = -1 < \sigma < 2$$

$$Y(s) = X(s)H(s) = \frac{-1}{(s-2)(s+1)} = \frac{-1/3}{s-2} + \frac{1/3}{s+1}$$

Based on the output ROC, take inverse Laplace transform

$$y(t) = \frac{1}{3}e^{2t}u(-t) + \frac{1}{3}e^{-t}u(t)$$

# **Distractor Logic**

Option A: Correct Answer

Option B: In the partial fraction if we feel pole '2' is right sided & pole '-1' is left sided

Option C: In the partial fraction expansion if we take negative sign of Y(s) as it is

Option D: In the partial fraction expansion if we miss  $\frac{1}{2}$  multipliers

24. The sound from two sound sources are 40 dB and 50 dB. Then the total sound intensity level in dB

Since 1995

Option A: 50.41

Option B: 52.39

Option C: 0

Option D:  $110 \times 10^3$ 

Ans: (A)

**Sol:** 40 dB = 
$$10 \log_{10} \left( \frac{I_1}{I_0} \right)$$
  $I_1 = 10^4 I_0 \left( \frac{W}{m^2} \right)$ 

50 dB = 
$$10 \log_{10} \left( \frac{I_2}{I_0} \right)$$
  $I_2 = 10^5 I_0 \left( \frac{W}{m^2} \right)$ 

Total sound intensity 
$$I_T = I_1 + I_2 = (10^4 + 10^5) I_0 \left(\frac{W}{m^2}\right)$$

Total sound intensity level (dB) =  $10 \log_{10} \left( \frac{1.1 \times 10^5 I_0}{J_{\odot}} \right) = 50.41 \text{ dB}$ 



### Distractor Logic

Option A: This is correct option.

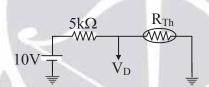
Option B: If we taken sound in dB =  $20 \log_{10} \left( \frac{I_1}{I_0} \right)$  as formula then total sound intensity level in dB = 52.39 dB.

Option C: Total sound intensity level is 0 dB if and only if  $I_T = I_0$  but in given a problem  $I_T \neq I_0$ .

Option D: This option is in magnitude form but as per question we want answer in dB format.

25. A thermistor sensor is used to monitor room temperature. It has resistance of 5 k $\Omega$  at 30 °C with a slope of -10 %/°C. The dissipation constant  $P_D = 5 \text{ mW/}$ °C. It is proposed to use the thermistor in provide V<sub>o</sub> voltage of 5.0 V at 30 °C. The value of the divider network shown in figure to thermistor resistance in  $k\Omega$  due to self heating effects is

Since 1995



Option A: 4.5

Option B: 5.5

Option C: 5

Option D: None of the above

Ans: (A)

**Sol:** At 30 °C 
$$R_{th} = 5k\Omega$$
  $V_D = \frac{R_{th}}{R_{th} + 5k\Omega} \times 10V = 5V$ 

Power dissipated in thermistor  $\Rightarrow P = \frac{V^2}{R_{th}} = \frac{(5)^2}{5 k} = 5 \text{ mW}$ 

$$\Delta T = \frac{P}{P_D} = \frac{5 \text{ mW}}{5 \text{ mW}/^{\circ} \text{ C}} = 1^{\circ} \text{C}$$

$$R_{th} = 5k\Omega + \left[1^{\circ}C \times \frac{-10}{100^{\circ}C} \times 5k\Omega\right] = 4.5 \text{ k}\Omega$$



# **Distractor Logic**

Option A: This is correct option

Option B: This option is correct if we take given slope value positive i.e. +10%/°C

Option C: This option is correct without considering self heating effect

Option D: None of the above



# **NEW BATCHES FOR**

# ESE - 2017 Stage - II (Mains)

BATCH - 1	BATCH - 2				
18 <sup>th</sup> Jan 2017	9 <sup>th</sup> Feb 2017 (E&T & ME)				
(E&T, EE, CE & ME)	15 <sup>th</sup> Feb 2017 (EE & CE)				

ESE - 2017 MAINS OFFLINE TEST SERIES
WILL BE CONDUCTED FROM MARCH 1<sup>ST</sup> WEEK
DETAILED SCHEDULE WILL BE ANNOUNCED SOON



An Iron-constantan thermocouple, having linear sensitivity of 20 μV/°C, is calibrated at cold junction temperature of 0°C. This type of thermocouple is used to measure an unknown temperature with the cold junction temperature of 25°C. Actual hot junction temperature in °C if the emf generated is 2.0 mV is

Option A: 125

Option B: 100

Option C: 25

Option D: None of the above

Ans: (A)

**Sol:** 
$$E_{0 \to T} = E_{0 \to 25^{\circ}C} + E_{25^{\circ}C \to T}$$

$$E_{25^{\circ}C \to T} = 2.0 \text{ mV}$$
  $E_{0^{\circ}C \to 25^{\circ}C} = 20 \times 10^{-6} \times 25 = 0.5 \text{ mV}$ 

$$E_{0\to T} = 0.5 \text{mV} + 2.0 \text{ mV} = 2.5 \text{ mV}$$

Actual hot junction temperature = 
$$\frac{2.5 \times 10^{-3} \text{ V}}{20 \times 10^{-6} \text{ V/}^{\circ} \text{ C}} = 125 \text{ }^{\circ}\text{C}$$

# **Distractor Logic:**

Option A: This is correct option

Option B: This option is correct if we take  $E_{0^{\circ}C \to T} = E_{25^{\circ}C \to T} = 2.0 \text{ mV}$ 

Option C: This option is correct if we take  $E_{0^{\circ}C \to T} = E_{0^{\circ}C \to 25^{\circ}C} = 0.5 \text{ mV}$ 

Option D: None of the above

Which of the following could not be the auto correlation function of a random process? 27.

Option A: 
$$R(\tau) = \begin{cases} 1 - |\tau|; & |\tau| < 1 \\ 0; & |\tau| > 1 \end{cases}$$

Option B:  $R(\tau) = 5\sin 3\tau$ 

Option C: 
$$R(\tau) = \begin{cases} \cos \tau; & |\tau| \le \pi/2 \\ 0; & |\tau| > \pi/2 \end{cases}$$

Option D: 
$$R(\tau) = \frac{\sin \tau}{\tau}$$



Ans: (B)

**Sol:** Auto correlation function is always even function, maximum value occurs at  $\tau = 0$ . option (B) violating these two conditions

# **Distractor Logic:**

Option A: ACF satisfies  $R_x(-\tau) = R_x(\tau)$ 

$$|R_x(\tau)| \le R_x(0)$$

$$R(\tau) = \begin{cases} 1 - |\tau|; & |\tau| < 1 \\ 0; & |\tau| > 1 \end{cases}$$
 satisfies both properties

So option A wrong

Option B:  $R_x(-\tau) \neq R_x(\tau)$ ,  $5\sin(3\tau)$  violates even condition So option (B) Correct Answer

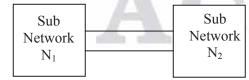
Option C:  $\cos(\tau)$ ,  $|\tau| \le \pi/2$  satisfies all properties

So option C is wrong

Option D:  $\frac{\sin \tau}{\tau}$  satisfies all properties

So option D is wrongs

Since 1995 A network is composed of two sub-networks N<sub>1</sub> & N<sub>2</sub> as shown in figure 28.



If the sub network N<sub>1</sub> contains only linear, bilateral, time invariant elements then it can be replaced by its Thevenin's equivalent even if the sub-network N<sub>2</sub> contains

Option A: A two-terminal element which is non linear

Option B: A non-linear inductance mutually coupled to an element in N<sub>1</sub>

Option C: An element which is linear, but mutually coupled to same element in N<sub>1</sub>

Option D: A dependent source the value of which depends upon the voltage (OR) current in any element of N<sub>1</sub>



Ans: (A)

Sol: Network N<sub>1</sub> is passive network as per given data. Then network N<sub>2</sub> should be active element which should be independent. i.e., independent either voltage (OR) current source

### **Distractor Logic**

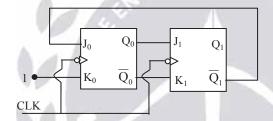
Option A: Correct option

Option B: Nonlinear inductance is not possible

Option C: If N<sub>2</sub> is linear element we can't replace the network by its Thevenin's equivalent

Option D: If N<sub>2</sub> contains dependent source then control variable should be in network N<sub>2</sub> only

Find the value of the following counter after 730 clock pulses. Initially  $Q_0 = Q_1 = 0$ . 29.



Option A:  $Q_0 Q_1 = 00$ 

Option B:  $Q_0 Q_1 = 01$ 

Option C: $Q_0 Q_1 = 10$ 

Option D:  $Q_0 Q_1 = 11$ 

Ans: (C)

Sol:

It is a 2 bit synchronous counter. Given  $J_0 = \overline{Q}_1$ ;  $K_0 = 1$ ;  $J_1 = Q_0$ ;  $K_1 = \overline{Q}_0$ 

	Prese	Fli	p Flo	p In	Nex				
	$Q_0$	$Q_1$	$J_0$	$K_0$	$J_1$	$K_1$	$Q_0$	$Q_1$	
0	0	0	1	1	0	1	1	0	2
2	1	0	1	1	1	0	0	1	1
1	0	1	0	1	0	1	0	0	0

Since 1995

counting sequence is 00, 10, 01, 00,..... it is a Mod-3 counter.

Counter value after 730 clock pulses is same value of the counter after 1 pulse i.e.,  $Q_0Q_1 = 10$ 



### **Distractor Logic**

Option A: It might be mistaken as 2-bit Johnson counter which is a 4:1 counter. Thus value after 730 pulses is same as initial value i.e.,  $Q_0 Q_1 = 00$ 

Option B: The counting sequence of the counter may be mistaken as 00, 01, 10, 00, ........ Then its solution is taken as 01

Option C: Correct option

Option D: Mistakenly chosen as 11

30. In 8085 Microprocessor the Accumulator has a 2's complement number '11110100'. Determine the function of executing the following sequence of instructions RLC, RRC and RAR.

Assume 'carry' flag is cleared initially.

Option A: It finds 1's complement representation of the given 2's complement number

Option B: It converts the given 2's complement number to corresponding sign magnitude representation

Option C: It divides the given 2's complement number by 2

Option D: It multiplies the given 2's complement number by 2

Ans: (C)

**Sol:** Given Accumulator value =  $-12_{10}$ 

		ACC						Si	cy	
		1	1	1	1	0	1	0	0	0
RLC	$\Rightarrow$	1	1	1	0	1	0	0	1	1
RRC	$\Rightarrow$	1	1	1	1	0	1	0	0	1
RAR	$\Rightarrow$	1	1	1	1	1	0	1	0	0

Accumulator value is  $11111010 = -6_{10}$ 

# Distractor Logic

Option A: It may be mistaken as 1's complement form of given number

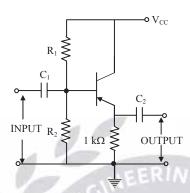
Option B: It may be mistaken as sign magnitude form of given number

Option C: Correct option

Option D: It may be mistaken as multiplication of 2's complement number



31. In the amplifier circuit shown in figure, the transistor parameters with usual notations are  $g_m = 0.015~S$ ,  $r_{b'e} = 1~k\Omega$ ,  $r_{bb'} = 90~\Omega$ ,  $C_{b'e} = 20~pF$  and  $C_{b'e} = 3~pF$ . Neglecting the loading effect of biasing resistors,  $R_1$  &  $R_2$ , the mid-frequency voltage gain of the amplifier is \_\_\_\_\_.



$$(A) - 13.76$$

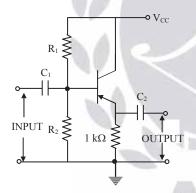
(B) 0.9411

(C) 0.936

(D)

Ans: (C)

Sol:



Consider the parameters of BJT

$$r_{b'e} = 1 k\Omega$$
,  $r_{bb'} = 90\Omega$  &  $g_m = 0.015$  S,

Step(1): 
$$h_{ie} = r_{b'e} + r_{bb'} = 1090 \Omega$$

Consider, 
$$r_{b'e} = \frac{h_{fe}}{g_m} \Rightarrow h_{fe} = r_{b'e} \times g_m = 1 \text{ k}\Omega \times 0.015 \text{ C} = 15$$

Step(2): The given circuit is emitter follower (CC Amplifier)

... The mid-frequency voltage gain in a CC amplifier,

$$A_{V} = \frac{(1 + h_{fe})R_{E}}{h_{ie} + (1 + h_{fe})R_{E}} = \frac{16 \times 1 k\Omega}{1.09 k\Omega + 16 \times 1 k\Omega} = \frac{16 k\Omega}{17.09 k\Omega}$$

$$A_V = 0.936$$



# **Distractor Logic**

Option: A:

If the given circuit is assumed as CE Amplifier, with a load resistance of R<sub>L</sub>=1k; then,

Step (1): 
$$h_{ie} = r_{b'e} + r_{bb'} = 1090 \Omega$$

Consider, 
$$r_{b'e} = \frac{h_{fe}}{g_m} \Rightarrow h_{fe} = r_{b'e} \times g_m = 1 \text{ k}\Omega \times 0.015 \text{ U} = 15$$

Step (2): 
$$A_V = -\frac{h_{fe}R_L}{h_{ie}} = -\frac{15 \times 1k}{1.09k} = -13.76$$

Option: B

**CC** Amplifier: if we consider  $h_{ie} = r_{b'e}$  [neglecting  $r_{bb'}$  (base spreading resistance)]

$$A_{V} = \frac{(1 + h_{fe})R_{E}}{h_{ie} + (1 + h_{fe})R_{E}} = \frac{16 \times 1k\Omega}{1k\Omega + 16 \times 1k\Omega} = \frac{16k\Omega}{17k\Omega} = 0.9411$$

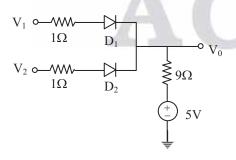
Option: C

CC Amplifier;  $A_V = 0.936$ 

Option: D

The voltage gain in a CC amplifier ideally is '1'

32. Assume both the diodes to be precise, if  $V_1 = -5V \& V_2 = 10V$ , find  $V_0$ ?



Option A: -4V

Option B: 8.87 V

Option C: 9.5 V

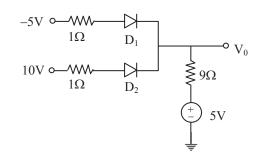
Option D: -4.63 V



Ans: (C)

**Sol:** Assume D<sub>1</sub> OFF & D<sub>2</sub> ON,

$$\therefore V_0 = \frac{\left(\frac{10}{1}\right) + \left(\frac{5}{9}\right)}{\left(\frac{1}{1}\right) + \left(\frac{1}{9}\right)} = 9.5V$$



Now, check Assumption:

$$i_{D_2} = \frac{10 - 9.5}{1} = 0.5A > 0,$$

$$\therefore$$
 D<sub>2</sub>  $\rightarrow$  ON  $\rightarrow$  True.

$$V_{D_1} = V_1 - V_0 = -5 - 9.5 = -14.5V < 0$$

 $\therefore$  D<sub>1</sub>  $\rightarrow$  OFF is True

# **Distractor Logic:**

Option A:  $-4 \text{ V} \rightarrow \text{This}$  is possible only when  $D_1 \rightarrow \text{short } \& D_2 \rightarrow \text{open}$ But from given biasing, this result is impossible

Option B: 8.87 V  $\rightarrow$  is possible for  $D_1 \rightarrow$  OFF &  $D_2 \rightarrow$  ON, but if diodes are practical. But given diodes are precise ⇒ They are ideal

Option C: True

Option D:  $-4.63 \text{ V} \rightarrow \text{This}$  is possible only when  $D_1 \rightarrow \text{short}$ ,  $D_2 \rightarrow \text{open } \& \text{ diodes are practical}$ 

33. The general solution of 
$$\frac{dy^4}{dx^4} - 6\frac{dy^3}{dx^3} + 12\frac{dy^2}{dx^2} - 8\frac{dy}{dx} = 0$$
 is

(A) 
$$y = C_1 + (C_2 + C_3 x + C_4 x^2) e^{2x}$$

(B) 
$$y = (C_1 + C_2 x + C_3 x^2) e^{2x}$$

(C) 
$$y = (C_1 + C_2 x + C_3 x^2 + C_4 x^3) e^{2x}$$

(D) 
$$y = C_1 + C_2 x + C_3 x^2 + C_4 e^{2x}$$

33. Ans: (A)

**Sol:** The given equation is  $(D^4 - 6D^3 + 12D^2 - 8D) y = 0$ 

$$D(D^3 - 6D^2 + 12D - 8) y = 0$$

$$D(D-2)^3 = 0$$

$$\therefore$$
 D = 0, 2, 2, 2

:. The required solution is (A)



34. 
$$L\{e^{-2t}[1-u(t-1)]\} = ?$$

(A) 
$$\frac{1 - e^{(s+2)}}{(s+2)}$$

(B) 
$$\frac{1 + e^{(s+2)}}{(s+2)}$$

(C) 
$$\frac{1-e^{-(s+2)}}{(s+2)}$$

(D) 
$$\frac{1+e^{-(s+2)}}{(s+2)}$$

**Sol:** 
$$L\{e^{-2t}[1-u(t-1)]\} = L\{e^{-2t}g(t)\}$$
 (Where  $g(t) = [1-u(t-1)]$ )

$$=G(s+2)$$
  $\left(G(s)=L\{g(t)\}=\frac{1}{s}-\frac{e^{-s}}{s}\right)$ 

$$=\frac{1-e^{-(s+2)}}{(s+2)}$$

35. The closed loop transfer function of a unity feedback system is 
$$\frac{C(s)}{R(s)} = \left(\frac{30}{s^2 + 5s + 36}\right)$$
. The steady

state error due to a unit step input is

Option A: 
$$\frac{36}{66}$$

Option B: 
$$\frac{1}{6}$$

Option C: 
$$\frac{36}{30}$$

Option D: None

Ans: (B)

**Sol:** Given CLTF 
$$\frac{C(s)}{R(s)} = \frac{30}{s^2 + 5s + 36}$$

Get OLTF G(s) = 
$$\frac{30}{s^2 + 5s + 6}$$
, H(s) = 1

Steady state error for unit step input 
$$e_{ss} = \frac{1}{1+k} = \frac{1}{1+\frac{30}{6}}$$

$$e_{ss} = \frac{6}{36} = \frac{1}{6}$$



# **Distractor Logic:**

Option A: If the given transfer function is considered as a OLTF then  $e_{ss} = \frac{A}{1+k}$ 

$$e_{ss} = \frac{1}{1 + \frac{30}{36}} = \frac{36}{66}$$

Option B: Correct Option

Option C: If the given transfer function is considered as OLTF and taken  $e_{ss} = \frac{A}{k} = \frac{1}{\left(\frac{30}{36}\right)} = \frac{36}{30}$ 

Option D: If considered OLTF is unstable then  $e_{ss}$  is none (or)  $\infty$ 

# OUR ESE 2016 TOP 10 RANKERS IN ALL STREAMS









OF STUDENTS
IN TOP 10
ARE FROM
ACE
and many more...

**29 RANKS IN TOP 10 IN ESE-2016** 



# Q.36 – Q.65 carry two marks each.

Consider the filter  $H(z) = \frac{z+2}{z+0.5}$ . The input to this filter is  $x(n) = \cos(n\pi)$ . The phase delay of this system is \_\_\_\_\_ secs

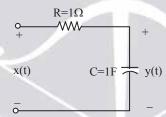
Ans: 1

**Sol:** 
$$\omega_0 = \pi(z = e^{j\omega_0} = -1) \Rightarrow H(e^{j\omega_0}) = H(e^{j\pi}) = H(-1) = \frac{-1+2}{-1+0.5} = -2 = 2e^{-j\pi}$$

$$x(n) = cos(n\pi) \Rightarrow y(n) = 2cos(n\pi - \pi)$$

$$t_{p}(\omega) = -\frac{\theta(\omega)}{\omega} = -\frac{(-\pi)}{\pi} = 1 \sec$$

Signal  $x(t) = |10 \sin t|$  Volts is applied to the circuit shown in figure. The Fourier series coefficient of the output has a dc value of



Ans: 6.366

**Sol:**  $x(t) = |10 \sin t|$  is a Full-wave rectifier

$$\omega_0 = 2 \operatorname{rad} / \sec$$
,

The Exponential Fourier Series coefficients of Full Wave rectified wave form is

$$C_0 = \frac{20}{\pi}$$
  $C_n = \frac{20}{\pi(1-4n^2)}$ 

$$H(\omega) = \frac{1}{1 + j\omega}$$

$$\Rightarrow H(n\omega_0) = \frac{1}{1 + jn\omega_0} = \frac{1}{1 + j2n}$$

$$\left. H(n\omega_0) \right|_{n=0} = 1$$

 $\therefore$  dc component of output is =  $C_0H(0) = \frac{20}{\pi} = 6.366$ 



When three transparent glass sheets of thickness 0.1 mm, 0.2 mm & 0.3 mm respectively are inserted in the interfering beam path in a Bi-prism experiment. The path difference due to insertion of three glass sheets in um is (Refractive index of glass sheet is 1.5)

Ans: 300

Sol: P.D = 
$$\mu(t_1 + t_2 + t_3) - (t_1 + t_2 + t_3)$$
  
=  $(\mu - 1)(t_1 + t_2 + t_3)$   
=  $(1.5 - 1)(0.1 + 0.2 + 0.3) \times 10^{-3}$   
=  $0.5 \times 0.6 \times 10^{-3}$   
=  $30 \times 10^{-5}$   
P.D =  $300 \mu m$ 

A dual slope integrating DVM is used to measure a voltage signal  $V(t) = (100+100\sin 100\pi t)V$ , in 39. its (0-200)V range of operation and  $3\frac{1}{2}$  digit display. The accuracy specification of this DVM is 0.5% of reading + 1 digit. The % error in reading of DVM is

Ans: 0.6% Range: 0.3% to 0.8%

**Sol:** DVM measures average value of input voltage signal.

$$\therefore$$
 DVM reading = 100 V

$$1 \text{ digit} = 1 \text{ count}$$

$$= 1 step$$

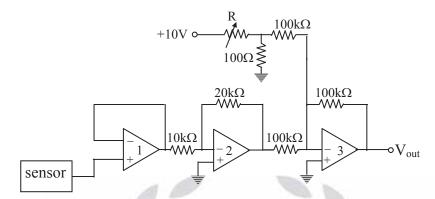
Resolution of DVM in 200V range is  $\frac{200}{2 \times 10^3} = 0.1 \text{ V}$ 

Error = 
$$\frac{0.5}{100} \times 100 + 0.1$$
  
= 0.6 V

% error in reading of 
$$100V = \frac{0.6}{100} \times 100\% = 0.6\%$$



40.



A sensor outputs a voltage ranging from -1 V to -6 V for interface to an analog to digital converter. The signal conditioning circuit output needs to be 0 V to 10 V. For given data the value of R in  $\Omega$  is

Ans: 400

Sol: Sensor Output Signal Conditioning Circuit Output

-1V

0V

-6V

10V

$$V_{out} = mV_{in} + V_{x}$$

$$0 = m(-1) + V_{x} ---- (1)$$

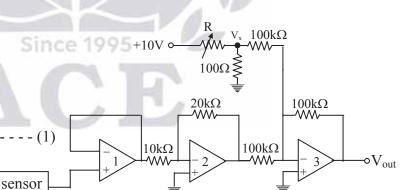
$$10 = m(-6) + V_{x}$$

$$10 = -5m$$

$$m = -2 \quad V_{x} = 2$$

$$V_{x} = 10 \times \frac{100k\Omega \parallel 100\Omega}{(100k\Omega \parallel 100\Omega) + R} ---- (1)$$

$$100 k\Omega \parallel 100 \Omega \cong 100 \Omega$$
sensor



Equation (1) becomes

$$V_x = 10 \times \frac{100}{100 + R}$$

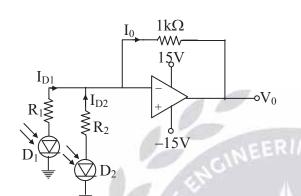
 $R = 400 \Omega$ 



Photodiodes  $D_1$  &  $D_2$  have active area of  $10^{-2}$  cm<sup>2</sup> &  $10^{-1}$  cm<sup>2</sup> respectively.  $D_1$  &  $D_2$  both have same sensitivity of 1 A/W. When the light of intensity 10 mW/cm<sup>2</sup> falls on the photodiode then the power dissipation in resistor 1 k $\Omega$  in mW is

Ans: 1.21 Range: 1.10 to 1.30

Sol:



$$\begin{split} &I_{0} = I_{D_{1}} + I_{D_{2}} \\ &I_{D_{1}} = 10 \frac{mW}{cm^{2}} \times 10^{-2} \text{ cm}^{2} \times 1 \frac{A}{W} = 0.1 \text{ mA} \\ &I_{D_{2}} = 10 \frac{mW}{cm^{2}} \times 10^{-1} \text{ cm}^{2} \times 1 \frac{A}{W} = 1 \text{ mA} \\ &I_{0} = 1.1 \text{ mA} \end{split}$$

A single phase house service energy meter of 230V, 10A. The meter constant of meter is  $180\left(\frac{\text{rev}}{\text{kWhr}}\right)$ . The meter takes 55sec for making 90 revolutions of the disc when connected to a 36

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kW unity power factor load. The error in the reading of the meter in % is

Ans: -9.09 Range: -8 to -10

**Sol:** Energy recorded (measured value)

 $P = I_0^2 \times 1000 = 1.21 \text{ mW}$ 

$$=\frac{90(\text{rev})}{180(\text{rev}/\text{kWhr})} = 0.5(\text{kWhr})$$



Energy consumed (true value )

$$= \frac{36 \,\text{kW} \times 55 \,\text{sec}}{3600} = 0.55 \big(\text{kWhr}\big)$$

$$Error = \frac{E_{measured} - E_{true}}{E_{true}} \times 100$$

$$= \frac{0.5 \text{ kWhr} - 0.55 \text{ kWhr}}{0.55 \text{ kWhr}} \times 100$$

$$= -9.09\%$$

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43. Ten different signals are to be Time Division Multiplexed and transmitted using PCM. Four of these signals have a maximum frequency of 10kHz, two of them have maximum frequency of 15kHz, two other signals have a maximum frequency of 5kHz and remaining signals have a maximum frequency of 20kHz. The value of bit rate (in Mbps), if signals sampled at Nyquist rate and samples are represented using 10 bits is \_\_\_\_\_\_

Ans: 2.4 Range: 2.3 to 2.5

**Sol:**  $r_{b,TDM} = n f_{s,TDM}$ 

$$n = 10$$

$$f_{s,TDM} = 4f_{s_1} + 2f_{s_2} + 2f_{s_3} + 2f_{s_4}$$

$$W_1 = 10k$$
,  $f_{s_1} = 20k$ 

$$W_2 = 15k$$
,  $f_{s_2} = 30k$ 

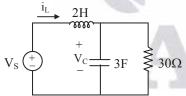
$$W_3 = 5k$$
,  $f_{s_2} = 10k$ 

$$W_4 = 20k$$
,  $f_{s_4} = 40k$ 

$$f_{s,\text{TDM}} = 4 \times 20k + 2 \times 30k + 2 \times 10k + 2 \times 40k = 240 \text{ k samples/sec}$$

$$r_{b,TDM} = 10 \times 240 \, k \, samples \, per \, sec = 2400 \, kbps = 2.4 \, Mbps$$

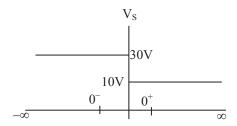
44. Consider the following network



If  $V_s = 10 + 20 \text{ u(-t)}$ , then the value of  $\frac{di_L(t)}{dt}$  at  $t = 0^+$  in A/s is \_\_\_\_\_

Ans: -10

Sol:



 $V_{\rm C}(0^-)_{\rm o}$ 

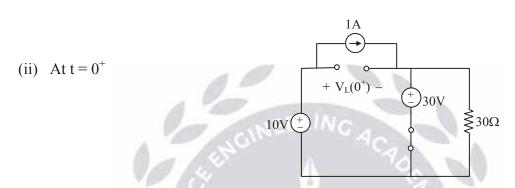
 $\leq 30\Omega$ 



(i) At  $t = 0^-$ : Steady state

$$i_L(0^-) = \frac{30}{30} = 1A = i_L(0^+)$$
 30V  $\stackrel{+}{=}$ 

$$V_C(0^-) = 30V = V_C(0^+)$$



At  $t = 0^+$ : Transient state

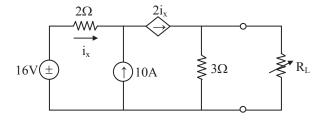
By KVL 
$$\Rightarrow 10 - V_L(0^+) - 30 = 0$$

$$\Rightarrow V_L(t)\Big|_{t=0^+} = -20$$

$$\Rightarrow \frac{Ldi_L(t)}{dt}\Big|_{t=0^+} = -20$$

$$\Rightarrow \frac{di_L(t)}{dt}\Big|_{t=0^+} = \frac{-20}{L} = \frac{-20}{2} = -10A/s$$

45. Consider the following network

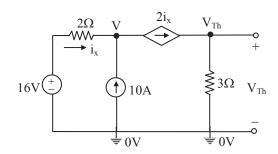


The value of maximum power transferred to load 'R<sub>L</sub>' (in watts) is \_\_\_\_\_\_

Ans: 300



**Sol:** Evaluation of  $V_{Th}$ :



Nodal 
$$\Rightarrow$$
  $-i_x - 10 + 2i_x = 0 \Rightarrow i_x = 10A$ 

Nodal 
$$\Rightarrow$$
  $-2i_x + \frac{V_{Th}}{3} = 0 \Rightarrow V_{Th} = 6i_x = 60V$ 

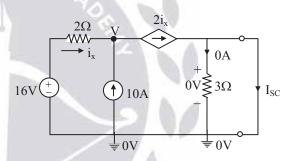
MCINEEUM

Evaluation of I<sub>SC</sub>:

Nodal 
$$\Rightarrow$$
  $-i_x - 10 + 2i_x = 0 \Rightarrow i_x = 10 \text{ A}$ 

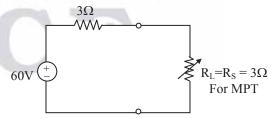
So, 
$$I_{SC} = 2i_x = 20A$$

$$\Rightarrow R_{Th} = \frac{V_{Th}}{I_{SC}} = \frac{60}{20} = 3\Omega$$



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$$\Rightarrow P_{\text{max}} = \frac{V_{\text{Th}}^2}{4 \times R_{\text{Th}}} = \frac{60^2}{4 \times 3} = 300 \text{W}$$



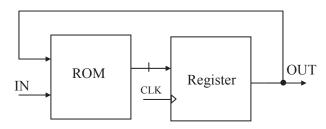
46. The maximum frequency of operation (in MHz) for the sequential circuit shown below is \_\_\_\_\_\_(MHz)

The Propagation delays are as follows

$$t_{ROM} = 7.5 \text{ ns}; \ t_{Reg} = 2.5 \text{ ns}$$

$$t_{Setup} = 2.5 \text{ ns}; t_{Hold} = 2.5 \text{ ns}$$





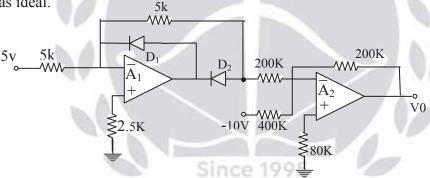
**Ans: 80** 

**Sol:** The frequency of operation of the given sequential circuit is

$$f \le \frac{1}{t_{Rom} + t_{Reg} + t_{Setup}} \Rightarrow f \le \frac{1}{(7.5 + 2.5 + 2.5) \times 10^{-9}}$$

$$f_{\text{max}} = \frac{1 \times 10^9}{12.5} = 80 \text{MHz}$$

Find the output voltage V<sub>0</sub> (in Volts) of op-amp circuit shown in figure assuming op-amp & diodes as ideal.

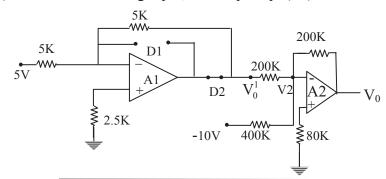


**Ans: 10** 

Sol: Step (1) For the given input of 5v, D1 is off & D2 is ON

$$V_0^1 = \frac{-5k}{5k} \times 5v = -5v$$

Step (2) KCL at the inverting input, V<sub>2</sub> of op-amp2(A<sub>2</sub>)





Op-amp 2 (A2)

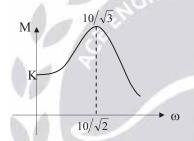
$$\frac{V_0^1}{200k} - \frac{10v}{400k} + \frac{V_0}{200k} = 0$$

$$V_0 = 200 k \left[ \frac{5V}{200k} + \frac{10V}{400k} \right] = 5 V + 5 V$$

$$\therefore$$
 V<sub>0</sub> =10 V

48. Frequency response of a second order system is given below. The value of K is \_\_\_\_\_

$$G(s) = \frac{K \times 100}{s^2 + 10s + 100}$$



Ans: 5

**Sol:** 
$$G(s) = \frac{K \times 100}{s^2 + 10s + 100}$$

$$s^2 + 10s + 100 \Longrightarrow s^2 + 2s\xi\omega_n + {\omega_n}^2$$

$$\omega_n = 10$$

$$\xi = 1/2$$

$$M = \frac{K}{2\xi\sqrt{1-\xi^2}} = \frac{10}{\sqrt{3}} = \frac{K}{2(0.5)\sqrt{1-(0.5)^2}} = \frac{10}{\sqrt{3}}$$

$$K = 5$$

49.  $y = e^{-x} \left( C_1 \cos \sqrt{3} x + C_2 \sin \sqrt{3} x \right) + C_3 e^{2x}$  is the general solution of

(A) 
$$(D^3 + 4) y = 0$$

(B) 
$$(D^3 - 8) y = 0$$

(C) 
$$(D^3 + 8) y = 0$$

(D) 
$$(D^3 - 2D^2 + D - 2) y = 0$$

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49. Ans: (B)



**Sol:** The roots of AE are  $\left(-1 \pm \sqrt{3} i\right)$  & 2

.. The required equation is

$$(D-2)(D^2+2D+4)y=0$$

i.e., 
$$(D^3 - 8) y = 0$$

- 50. Find the value of  $\int_{C} \frac{z \cos z}{\left(z \frac{\pi}{2}\right)^2} dz$ , where 'C' is |z 1| = 1
  - (A) iπ
- (B)  $-i\pi$
- (C)  $i\pi^2$
- (D)  $-i\pi^2$

50. Ans: (D)

**Sol:**  $z = \frac{\pi}{2} = \frac{3.14}{2} = 1.57$  is a pole of order '2' lies inside 'C'

$$\iint_{C} \frac{z \cos z}{\left(z - \frac{\pi}{2}\right)^{2}} dz = 2\pi i f^{1} \left(\frac{\pi}{2}\right) \quad \text{(where } f(z) = z \cos z\text{)}$$

$$= 2\pi i \left(\frac{-\pi}{2}\right)$$
$$= -\pi^2 i$$

51. 
$$(a\alpha + b) x + ay + bz = 0$$

$$(b\alpha + c) x + by + cz = 0$$

 $(a\alpha + b) y + (b\alpha + c) z = 0$  have non-trivial solutions if

- (i) a, b, c are in A. P
- (ii) a, b, c are in G. P
- (iii) a, b, c are in H. P
- (iv) ' $\alpha$ ' is a root of  $(ax^2 + 2bx + c) = 0$
- (A) both (i) & (iv)
- (B) both (ii) & (iv)
- (C) both (iii) & (iv)
- (D) only (iv)

51. Ans: (B)

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Sol: 
$$\begin{vmatrix} (a\alpha + b) & a & b \\ (b\alpha + c) & b & c \\ 0 & (a\alpha + b) & (b\alpha + c) \end{vmatrix} = 0$$

$$R_3 \rightarrow R_3 - (\alpha R_1 + R_2)$$

$$\Rightarrow \begin{vmatrix} a\alpha + b & a & b \\ b\alpha + c & b & c \\ -a\alpha^2 - 2b\alpha - c & 0 & 0 \end{vmatrix} = 0$$

i.e., 
$$-(a\alpha^2 + 2b\alpha + c)(ac - b^2) = 0$$

$$\therefore$$
 '\alpha' is a root of  $(ax^2 + 2bx + c) = 0$ 

or a, b, c are in G. P

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- Which one of the following statement is NOT TRUE about eigen signal?
  - Option A: Every signal is an eigen signal to the system described by  $h(t) = A\delta(t)$
  - Option B: The signal  $x(t) = e^{i\beta t}$  is an eigen signal of an LTI system given by  $h(t) = e^{-\alpha t}u(t)$
  - Option C: The signal  $x(t) = \cos \beta t$  is an eigen signal of an LTI system given by  $\dot{y}(t) + \alpha y(t) = x(t)$
- Option D: The signal  $x(t) = \operatorname{Sinc}(\alpha t)$  is an eigen signal of LTI system having  $h(t) = \operatorname{Sinc}(\beta t)$ ;  $\beta \ge \alpha$ Ans: (C)

**Sol:** If  $x(t) = \cos(\beta t) & \dot{y}(t) + \alpha y(t) = x(t)$  then

$$y(t) = A\cos(\beta t) + B\sin(\beta t)$$
$$= C\cos(\beta t + \theta)$$

### **Distractor Logic:**

Option A:  $x(t)*A\delta(t) = Ax(t)$  but you may feel  $A\delta(t)$ 

Option B: we may think of ' $\alpha$ ' & ' $\beta$ ' nature

Option C: Correct option

Option D: Wrong option

$$Sinc(t)*Sinc(t) = Sinc(t)$$

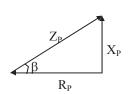
$$Sinc(t)*Sinc(2t) = \frac{1}{2}Sinc(t)$$
 but you may think  $Sinc(2t)$ 

An electrodynamometer wattmeter measures a power in single phase 50Hz AC circuit. Load voltage 230V and current 10A at lagging p.f of 0.5. The Wattmeter pressure coil resistance is  $10k\Omega$  and inductance is 100mH, then % error will be

$$(B) -1.68\%$$

Ans: (C)

Sol: 
$$\tan \beta = \frac{X_P}{R_P}$$
  
=  $\frac{2\pi \times 50 \times 100 \times 10^{-3}}{10 \times 10^3}$   
 $\beta = 0.179$ 





% error = + 
$$(\tan\phi \tan\beta) \times 100$$
  
= 1.732 ×0.00312 ×100  
= 0.54 %

## **Distractor Logic:**

Option: (A)

There is no effect of pressure coil resistance on wattemeter reading

Option: (B)

if it is energy meter

% error = 
$$\frac{\sin(10-30)-0.5}{0.5} \times 100$$
  
= -1.68%

Option: (C) Correct Answer

Option: (D)

% error = 
$$+ (\cos\phi \tan\beta) \times 100$$
  
=  $0.5 \times 0.00312 \times 100$   
=  $0.156\%$ 

The four arms of a Wheatstone bridge are as follows: Arm AB =  $100 \Omega$ , BC =  $10\Omega$ ,

CD =  $4\Omega$  and DA =  $50\Omega$ . The galvanometer has a resistance of  $20\Omega$  and is connected across BD.

A source of 10V DC is connected across AC, then the current through the galvanometer is

(A) zero

(B) 13.3mA

(C) 5.18mA

(D) 39mA

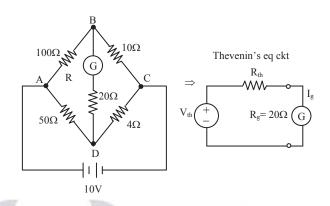
Ans: (C)

Sol:

$$V_{th} = V_B - V_D$$



$$\begin{split} &= \left(10 \times \frac{10}{10 + 100} - 10 \times \frac{4}{4 + 50}\right) \\ &= 0.9091 - 0.7407 \\ &= 0.17V \\ R_{th} &= \frac{100 \times 10}{100 + 10} + \frac{50 \times 4}{50 + 4} \\ &= 9.09091 + 3.7037 \\ &= 12.7946\Omega \\ I_{g} &= \frac{V_{th}}{R_{th} + 20} = \frac{0.17}{12.7946 + 20} = 5.18 \text{mA} \end{split}$$



# **Distractor Logic**

Option: (A)

if bridge is balanced,  $i_g = 0$ 

Option: (B)

if galvanometer resistance is neglected

$$0.17V \xrightarrow{R_{th}=12.79 \Omega} WW$$

$$i_g = \frac{0.17}{12.179} = 0.01329 \text{ A} = 13.3 \text{ mA}$$

**Option: (C)** Correct Answer

Option: (D)

if AB & AD branches are interchanged

$$V_{th} = \frac{10 \times 10}{10 + 50} - \frac{10 \times 4}{4 + 100} = 1.28 \text{ Volts}$$



$$R_{th} = \frac{10 \times 50}{10 + 50} + \frac{4 \times 100}{104} = 12.18 \ \Omega$$

$$i_g = \frac{1.28}{20 + 12.18} = 0.03977 \text{ A}$$

$$= 39 \text{ mA}$$

55. Piezoelectric transducer shown in figure has charge sensitivity of 2 pC/N. If a force of sin(t) N is applied to piezoelectric transducer the output amplitude magnitude of the charge amplifier in µV is



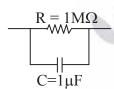
Option B: 
$$\frac{1}{\sqrt{2}}$$

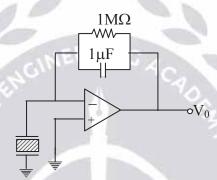
Option C: 2

Option D: 0



Sol:





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$$i = \frac{dq}{dt} = \frac{d}{dt}(dF) = \frac{d}{dt}(2 \times 10^{-12} \times \sin t) = 2\cos t \text{ (pA)}$$

$$Z = R \parallel \frac{1}{j\omega C} = \frac{R}{1 + j\omega RC} = \frac{10^6}{1 + j}$$

$$\left|Z\right| = \frac{10^6}{\sqrt{2}}$$

$$V_0 = -i \times Z = -2 \times \frac{10^6}{\sqrt{2}} \times 10^{-12} \text{ cost} = -\sqrt{2} \text{ cost } \mu \text{V}$$

$$|V_0| = \sqrt{2} \mu V$$



### Distractor Logic

Option A: This is the correct option

Option B: This option as answer if we take charge sensitivity = 1 pC/N

Option C: This option as answer if we take impedance  $Z = R = 1000 \Omega$ 

Option D: This option as answer if we apply a force of constant amplitude

56. Pre-envelope of  $cos(2\pi f_c t)$  is

Option A: 1

Option B:  $e^{j2\pi f_c t}$ 

Option C:  $e^{-j2\pi f_c t}$ 

Option D: 1/2

Ans: (B)

Sol: Pre-envelope, 
$$x_+(t) = x(t) + j\hat{x}(t) = \cos(2\pi f_c t) + j\sin(2\pi f_c t)$$
  
=  $e^{j2\pi f_c t}$ 

## Distractor Logic

Option A: Envelope of  $cos(2\pi f_c t)$  is 1. [If student think pre envelope and envelope are same]

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So option A wrong

Option B: Correct answer

Option C: 
$$x_+(t) = x(t) - j\hat{x}(t)$$
 [ wrong expression ]  
=  $\cos(2\pi f_c t) - j\sin(2\pi f_c t)$   
=  $e^{-j2\pi f_c t}$ 

So option C wrong

Option D: If student thinks

Pre envelope = 
$$\frac{1}{2}$$
 Envelope

So option D wrong



A signal is given by,  $s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$ . How many bits of quantization are required so that the signal to quantization noise ratio is greater than 50dB?

Option A: 8

Option B: 7

Option C: 9

Option D: 10

Ans: (C)

**Sol:** 
$$s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$$

Signal Power = 
$$\frac{20^2}{2} + \frac{17^2}{2} = 344.5$$

 $10\log SQNR \ge 50 \text{ dB}$ 

$$SQNR \ge 10^5$$

SQNR ≥ 100000

$$\frac{344.5}{QNP} \ge 100000$$

$$QNP \le \frac{344.5}{100000}$$
 
$$QNP \le 3.445 \times 10^{-3}$$

ONP 
$$\leq 3.445 \times 10^{-3}$$

$$QNP = \frac{\Delta^2}{12} = \left(\frac{DR}{L}\right)^2 \times \frac{1}{12}$$

Dynamic range = 
$$V_{max} - V_{max} = 37 - (-37) = 74$$

$$\left(\frac{74}{L}\right)^2 \times \frac{1}{12} \le 3.445 \times 10^{-3}$$

$$\frac{1}{L^2} \le \frac{3.445 \times 10^{-3} \times 12}{74 \times 74}$$

$$\frac{1}{L^2} \le 7.55 \times 10^{-6}$$

$$L^2 \ge 132450.331$$

$$L \ge 363.93$$

$$n \ge log_2 363.93$$

$$n \ge 8.51 \Rightarrow n \ge 9$$



### **Distractor Logic**

Option A:  $s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$ 

Signal Power = 
$$\frac{20^2}{2}$$
 = 200 [: maximum amplitude signal consider)

$$10 \log SQNR \ge 50$$

$$SQNR \ge 10^5$$

$$\frac{200}{ONP} \ge 100000$$

$$\frac{\text{QNP}}{200} \le \frac{1}{100000}$$

$$QNP \le \frac{200}{100000} \le \frac{1}{500} \Longrightarrow \frac{\Delta^2}{12} \le \frac{1}{500}$$

$$\Rightarrow \Delta^2 = 0.024 \Rightarrow \left(\frac{DR}{L}\right)^2 \leq 0.024 \Rightarrow \left(\frac{40}{L}\right)^2 \leq 0.024 \Rightarrow \left(\frac{L}{40}\right)^2 \leq 41.67$$

$$L^2 \ge 66672$$

$$L \ge 258.21$$
,  $n = 8$ 

So option A wrong

# Option B:

$$s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$$

Signal Power = 
$$\frac{20^2}{2} + \frac{17^2}{2} = 344.5$$

$$SQNR \ge 10^5$$

$$\frac{344.5}{\text{ONP}} \ge 100000$$

$$QNP \leq 0.003445\,$$

$$\frac{\Delta^2}{12} \le 0.003445$$

$$\Delta^2 \le 0.04134$$



$$\left(\frac{DR}{L}\right)^2 \le 0.04134$$

$$\left(\frac{34}{L}\right)^2 \le 0.04134$$
 ('.' dynamic range wrong)

$$L^2 \ge 34^2 \times \frac{1}{0.04134}$$

$$L^2 \ge 27963.23174$$

$$L \ge 167.22$$

$$n \ge 7.38$$

So option B wrong

Option C: Correct answer

Option D: If dynamic range is 40, n = 10

So option D wrong

58. For an RL series circuit  $R=5\Omega$  and  $X_L=5\Omega$ , the applied voltage  $V(t)=2\cos 3t+4\sqrt{2}\cos (3t+45^\circ)+12\sin 3t \ Volts.$  Then find the RMS value of the resultant current in the circuit.

Option A: 
$$\frac{3}{\sqrt{5}}$$
 A

Option B: 1A

Option C:  $\sqrt{2}$  A

Option D: None

Ans: (B)

Sol: 
$$V(t) = 2\cos 3t + 4\sqrt{2}\cos(3t + 45^{\circ}) + 12\sin 3t$$
  
=  $2\cos 3t + 4\sqrt{2}\left(\cos 3t \frac{1}{\sqrt{2}} - \sin 3t \frac{1}{\sqrt{2}}\right) + 12\sin 3t$ 

 $= 6\cos 3t + 8\sin 3t$ 



$$V_{RMS} = \sqrt{\frac{1}{2}(6^2 + 8^2)} = \sqrt{50} = 5\sqrt{2} \text{ Volts}$$

$$\left|I_{RMS}\right| = \left|\frac{V_{RMS}}{Z}\right| = \frac{5\sqrt{2}}{\sqrt{R^2 + X_L^2}} = \frac{5\sqrt{2}}{\sqrt{5^2 + 5^2}} = 1Amps$$

### **Distractor Logic:**

Option A: 
$$V_{RMS} = \sqrt{\frac{1}{2}(2^2 + (4\sqrt{2})^2 + 12^2)} = \sqrt{90} = 3\sqrt{10} \text{ Volts}$$

$$\left|I_{RMS}\right| = \left|\frac{V_{RMS}}{Z}\right| = \frac{3\sqrt{10}}{\sqrt{50}} = \frac{3}{\sqrt{5}} Amps$$

Option B:  $I_{RMS} = 1A$ 

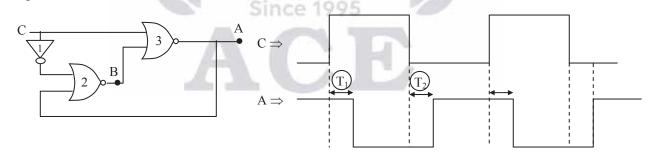
Option C:  $V = 6\cos 3t + 8\sin 3t = 10\sin(3t + 36.86^{\circ}) = 10\angle 36.86^{\circ}$ 

$$I = \frac{V}{Z} = \frac{10}{\sqrt{5^2 + 5^2}} = \frac{10}{\sqrt{50}} = \sqrt{2}Amps$$

Option D: None

59. In the following circuit 'C' input is driven by a square wave with 50% Duty cycle.

Each gate has a propagation delay of 1ns. Determine the values of  $T_1$  and  $T_2$ , respectively for the output waveform is as shown below.



Option A: 3ns, 3ns

Option B: 1ns, 3ns

Option C: 1ns, 2ns

Option D: 2ns, 3ns

Ans: (B)



**Sol:** When C = 1, A becomes 0 after 1ns.

When C = 0, B becomes 0 after 2ns and A is 1 after 3ns

#### **Distractor Logic**

Option A: There is a possibility of adding propagation delays of all logic gates

Option B: Correct option

Option C: It is possible to take Gate-3 delay when C = 1 and propagation delays of Gate② and Gate  $\Im$  when C = 0

Option D: Possible to take Gate  $\mathbb{Q}$ ,  $\mathbb{Q}$  delays when  $\mathbb{C} = 1$  and Gate  $\mathbb{Q}$ ,  $\mathbb{Q}$  and  $\mathbb{Q}$  delays when  $\mathbb{C} = 0$ 

In a 12-bit bipolar ADC, the digital output is in 2's complement form. Find the digital output for an 60. analog input of -4V in 4 Hex digits. The input voltage range is -4V to +4V

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Option A: FFFF<sub>H</sub>

Option B: 8000<sub>H</sub>

Option C: F800<sub>H</sub>

Option D: FF00<sub>H</sub>

Ans: (C)

**Sol:** For -4V, the 12-bit digital output is  $1000\ 0000\ 0000 = 800_{H}$ .

Then digital output in 4 Hex digits is =  $F800_H$ 

#### **Distractor Logic**

Option A:

It is possible to select max negative output as

1111 1111 1111 which is FFFF<sub>H</sub> in 4 Hex digits

Option B:

Max negative value using 12-bits is 1000 0000 0000<sub>2</sub>.

i.e.,  $800_{\rm H}$ . In 4 Hex digits it may be mistaken as  $8000_{\rm H}$ 

Option C:

Correct option

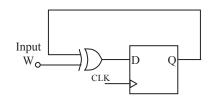
Option D:

Mistakenly it is taken as F00<sub>H</sub>, which is in 4 Hex digits as FF00<sub>H</sub>

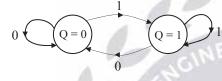
Since 1995



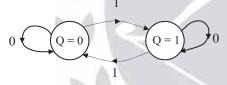
61. The state diagram for the sequential circuit shown below is



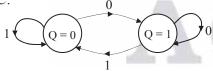
Option A:



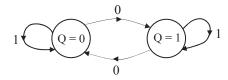
Option B:



Option C:



Option D:





**Ans: (B)** 

Sol:

$$Q(t+1) = D$$
 and output  $P = Q(t)$ 

Where 
$$D = Q \oplus W$$

i.e., 
$$Q(t+1) = Q(t) \oplus W$$

If 
$$W = 0 \Rightarrow Q(t+1) = Q(t)$$

If 
$$W = 1 \Rightarrow Q(t+1) = \overline{Q(t)}$$

### **Distractor Logic**

Option A:

The output branches from each state are not assessed properly

Option B:

Correct option

Option C:

The output branches are not assessed properly

Option D:

The output branches from each state are not assessed properly

- In the circuit shown in figure , a silicon transistor with  $V_{BE}$  = 0.7V,  $\beta$  = 100 is used. Then find the 62. collector current I<sub>C</sub>.
  - (A) 9.2 mA
- (B) 2.1 mA
- (C) 1.86 mA
- (D) 0

Ans: (B)

Step (1): KVL for BE loop of BJT
$$0 - 0.7V - I_E 1K + 10V = 0 \dots (1)$$

$$\Rightarrow I_E = \frac{9.3V}{1K} = 9.3\text{mA} \dots (2) \text{ [i.e J}_E \text{ is FB]}$$

$$\Rightarrow I_C = \left(\frac{\beta}{1+\beta}\right)I_E = 9.2\text{mA} \dots (3)$$



Step (2): KVL for C-loop

$$10V - I_C \times 5K - V_C = 0 \dots (4)$$

$$V_C = 10V - 9.2mA \times 5K = -36V \dots (5)$$

$$\Rightarrow$$
 V<sub>CB</sub> = V<sub>C</sub> - V<sub>B</sub> = -36V - 0 = -36V ..... (6)

NOTE:  $V_{CB}$  is  $-V_{CB}$  in  $-V_{CB}$  is  $-V_{CB}$  in  $-V_{CB}$  is  $-V_{CB}$  in  $-V_{CB}$  in  $-V_{CB}$  in  $-V_{CB}$  in  $-V_{CB}$  is  $-V_{CB}$  in  $-V_{CB}$  in

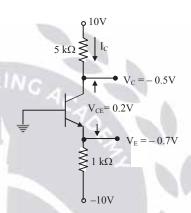
:. BJT is operated in saturation region

Step (3):  $\dot{}$  BJT is in saturation,  $V_{CE_{sat}} = 0.2V$ 

$$\Rightarrow$$
 V<sub>C</sub> = V<sub>CE<sub>cot</sub></sub> + V<sub>E</sub> = -0.5V ..... (1)

KVL for collector –loop:

$$I_C = \frac{10V - (-0.5V)}{5K} = 2.1 \text{mA} \dots (2)$$



# **Distractor Logic**

Option: A

If the device(BJT) is in forward active region, KVL for BE loop of BJT

$$0 - 0.7V - I_E 1K + 10V = 0$$

$$I_E = \frac{9.3V}{1K} = 9.3mA$$

$$\therefore I_{c} = \left(\frac{\beta}{1+\beta}\right) I_{E} = \frac{100}{101} \times 9.3 \text{mA}$$

$$I_c = 9.2 \text{mA}$$

Option: B

 $I_c = 2.1 \text{mA} \dots (1)$  (: Device is actually biased in saturation region)

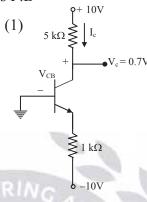


### Option: C

If the device is in inverse (or reverse) active region,

(i.e) E-B junction is R.B & C.B junction is F.B

Assuming 
$$V_{CB} = 0.7V \Rightarrow V_c = 0.7V \dots (1)$$



$$[::V_{CB}=V_{C}-V_{B}=V_{C}]$$

KVL for collector loop of BJT

$$10V - I_c 5K - 0.7V = 0 \dots (1)$$

$$I_c = \frac{9.3V}{5K} = 1.86\text{mA}....(2)$$

## Option: D

If the device is in cutoff region

$$I_B = 0 \Rightarrow I_C = 0 \dots (1)$$

- A bag contains (n + 1) coins. It is known that one of these coins shows heads on both sides where as the other coins are fair. One coin is selected at random and tossed. If the probability that the toss results in heads is  $\frac{7}{12}$ , then the value of 'n' is
  - (A) 4
- (B)5

- 63. Ans: (B)
- **Sol:** Let A = Event of selecting two headed coin

B = Event of selecting a fair coin

E = Event of coming head on selected coin

Given that  $P(A) P(E/A) + P(B) \cdot P(E/B) = \frac{7}{12}$ 

i.e., 
$$\frac{1}{(n+1)} \cdot 1 + \frac{n}{(n+1)} \cdot \frac{1}{2} = \frac{7}{12}$$

$$12 + 6n = 7n + 7$$

$$\therefore$$
 n = 5



The characteristic equation of a feedback control system is given by  $s^2 + s(k-1) + k = 0$ . Where k > 0. The break points on RLD are

Option A: -0.414, 2.414

Option B: 0.414, 2.414

Option C: 0.414, -2.414

Option D: -0.414, -2.414

**Ans: (C)** 

**Sol:** Break point 
$$\Rightarrow \frac{dk}{ds} = 0$$

Given, CE:  $s^2 + sk - s + k = 0$ 

$$CE \rightarrow (s^2-s)+k(s+1) = 0$$

$$k = -\frac{(s^2 - s)}{(s+1)}$$

$$\frac{dk}{ds} = - \left[ \frac{(2s-1)(s+1) - (s^2 - s)}{(s+1)^2} \right] = 0$$

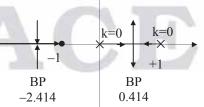
$$\Rightarrow 2s^2 - s + 2s - 1 - s^2 + s = 0$$

$$\Rightarrow s^2 + 2s - 1 = 0$$

$$s = \frac{-2 \pm \sqrt{4 - 4(1)(-1)}}{2} = \frac{-2 \pm \sqrt{8}}{2}$$

s = 0.414 and -2.414

$$\Rightarrow$$
 G(s)H(s) =  $\frac{k(s+1)}{s(s-1)}$   $\Rightarrow$ 



# **Distractor Logic**

Option A: Given points are not on RLD.

Option B: One BP is valid and another is invalid.

Option C: Correct Option.

Option D: One BP is valid and another is invalid.



The state model of the system is given as

$$\overset{\bullet}{\mathbf{X}} = \begin{bmatrix} 0 & 1 \\ -2 & -9 \end{bmatrix} \mathbf{X} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \mathbf{U} & \mathbf{\hat{x}} & \mathbf{\hat{y}} = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{\hat{x}}$$

Then find the poles of the system.

Option A: s = -0.22, -8.77

Option B: s = 0.216, -9.216

Option C: s = 0.22, 8.77

Option D: s = -0.216, 9.21

Ans: (A)

**Sol:** From the magnitude equation

$$[sI - A] = \begin{bmatrix} s & -1 \\ 2 & s + 9 \end{bmatrix}$$

$$Adj[sI - A] = \begin{bmatrix} s+9 & 1 \\ -2 & s \end{bmatrix}$$

$$TF = \frac{C \operatorname{Adj}[sI - A]B}{|sI - A|} + \mathcal{D}^{0}$$

$$TF = \frac{\begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} s+9 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ -2 & s \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix}}{s(s+9)+2} = \frac{\begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ s \end{bmatrix}}{s^2+9s+2} = \frac{1}{s^2+9s+2}$$

CE 
$$s^2 + 9s + 2 = 0$$

$$s = -0.22, -5.77$$

# **Distractor Logic**

Option A: Correct option

constructed with wrong calculation Option B: This option is determinant as  $|sI - A| = s^2 + 9s - 2 = 0$ 

Option C: While calculating determinant, the subtraction of last row and column element is wrong, then the equation is  $|sI - A| = s^2 - 9s + 2 = 0$ 

Option D: This option is constructed with calculation wrong of determinant as  $|sI - A| = s^2 - 9s - 2 = 0$