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Branch: ECE

ACE Pre-GATE 2017

	Q	0.1 – Q.5 Carry One Mark Each				
01.	Reaching a place of appointme	ent on Friday. I found that I was two days earlier than the scheduled				
	day. If I had reached on the following Wednesday then how many days late would I have been?					
	(a) one (b) Two					
	(c) three	(d) four				
01.	Ans: (c)					
Sol:	Friday \rightarrow 2 days earlier					
	Therefore, scheduled day = Friday + 2 = Sunday					
	Sunday + 3 = Wednesday					
	Therefore, I would have been late by 3 days					
	E 0 V					
02.	Choose the most appropriate	phrase from the options given below to complete the following				
	sentence.	Since 1995				
	The bus stopped to	more passengers.				
	(a) Take in	(b) Take on				
	(c) Take up	(d) Take for				
02.	Ans: (b)					
03.	Choose the appropriate sentence from the following options.					
	(a) She has been discharged since.					
	(b) She has since been discharged.					
	(c) She has been since discharge	ged.				
	(d) She since has been discharge	ged.				
03.	Ans: (b)					

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04.	Fill in the blank with an appropriate phrase.					
	The jet into the air.					
	(a) Soared.	(b) Soured.				
	(c) Sourced.	(d) Sored.				
04.	Ans: (a)					
05.	Choose the most appro	opriate word from the opti	ions given below to complete the following	g		
	If I had known that you v	were coming, I	_ you at the airport.			
		All American commences				
	(a) Would meet	(b) Would have met				
	(a) Would meet(c) Will have met	(b) Would have met(d) Had met	4cq			
05.	` '		ACADEM			

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Q.6 - Q.10 Carry two marks each

Which of the following can be logically inferred from the given statement. 06.

"No other studied medicine except Helen"

- (a) Helen only studied medicine
- (b) Only Helen studied medicine
- (c) Helen studied only medicine
- (d) Helen studied medicine only

06. Ans: (b)

The average electricity bill of a household for January to June is ₹ 980, for July to September is 07. ₹ 670, for October to December is ₹ 720. If the family goes on vacation for June and July and no electricity is used, what would be the average electricity bill for that year?

- (a) ₹ 500
- (b) ₹ 600
- (c) ₹ 700
- (d) ₹ 800

07. Ans: (c)

Sol: Average electricity bill from January to June = ₹ 980

∴ Total electricity bill from January to May = $980 \times 5 = ₹4900$

(As no electricity is used in June)

Similarly, total electricity bill from August to September (as no electricity is used in July)

And total electricity bill from October to December = $720 \times 3 = 2160$

Therefore, total electricity bill from January to December = 4900 + 1340 + 2160 = ₹8400

Thus, average electricity bill for the whole year = $\frac{8400}{12}$ = ₹700

08. The following question has four statements of three segments each. Choose the alternative where the third segment in the statement can be deduced using both the preceding two but not just from one of them.

A. Sonia is an actress. Some actresses are pretty. Sonia is pretty.

B. All actors are pretty. Manoj is not an actor. Manoj is not pretty



C. Some men are cops. Some men are brave. Some brave people are cops.

D. All cops are brave. Some men are cops. Some men are brave.

(a) only C

(b) only A

(c) only D

(d) B and C

08. Ans: (c)

Sol: Statements:

All cops are brave

Some men are cops



Conclusion:

Some men are brave (True)

Hence, only D follows.

09. A contractor, who got the contract for building the flyover, failed to construct the flyover in the specified time and was supposed to pay ₹ 50,000 for the first day of extra time. This amount increased by ₹ 4,000 each day. If he completes the flyover after one month of stipulated time, he suffers a loss of 10% in the business. What is the amount he received for making the flyover in crores of rupee? (One month = 30 days)

(a) 3.1

(b)3.24

(c) 3.46

(d) 3.68

09. Ans: (b)

Sol: The sum of money that the contractor was supposed to pay for the period of an month over the stipulated time is

$$=S_n = \frac{n}{2}[2a + (n-1)d]$$

$$a = 50,000, n = 30, d = 4000$$



$$S_{30} = \frac{30}{2} [2 \times 50,000 + (30 - 1) \times 4000] = 15[100,000 + 29 \times 4000]$$

₹ 3240000 = ₹ 32.4 lakhs

Loss in the business = 10%

- ∴ Amount he received for making the flyover = $\frac{3240000}{0.1}$ = 32400,000 = ₹ 3.24 crores
- 10. Study the following pie chart and table carefully to answer the following question that follow: Percentage break up of employees working in various departments of an organisation and the ratio of men to women in them.

Total number of employees = 1800

Percentage break up of employees:



Ratio of men to women

Department	Men	Women
Production	11	1
HR	1	3
IT	5	4
Marketing	7	5
Accounts	2	7

What is the number of men working in the marketing department?

- (a) 132
- (b) 174
- (c) 126
- (d) 189

10. Ans: (d)

Sol: Number of men working in the marketing department = $1800 \times \frac{18}{100} \times \frac{7}{12} = 189$



Q.11 – Q.35 Carry one mark each.

- 11. The signal $x(t) = \cos(50\pi t) + \cos(80\pi t)$ is sampled at 200Hz. The minimum number of samples required to prevent leakage is
- 11. **Ans: 40**

Sol:
$$\frac{\omega_1}{2\pi} = \frac{50\pi/200}{2\pi} = \frac{1}{8} = \frac{5}{40}$$

$$\frac{\omega_2}{2\pi} = \frac{80\pi/200}{2\pi} = \frac{1}{5} = \frac{8}{40}$$

- :. The minimum number of samples required to prevent leakage is 40.
- An n-channel JFET has a gate cut-off voltage of -10V. If -1V is applied at the gate terminal the minimum drain to source voltage required to obtain maximum drain resistance is _____(V) (Neglect channel length modulation)
- 12. Ans: 9
- **Sol:** Drain resistance will be maximum at saturation

Minimum voltage (V_{DS}) for saturation = $V_{GS} - V_P$

Given $V_P = Gate cut-off voltage = -10 V$

$$V_{DS} = V_{GS} - V_P = -1 - (-10) = +9 \text{ V}$$

A fixed radar, used for navigational purpose, is operating at 3 GHz and transmitting 100 kW of power. If the smallest ocean going ship has a radar cross section of 200 m² and the radar antenna gain is 15dB, then the effective range of the radar is (in km), if detection requires a minimum detectable signal power density of 1 nW/m² at the radar antenna.

Since 1995

Ans: 7.9 Range: 7.8 to 8.1

Sol: Given: $10\log G = 15 \Rightarrow G = 10^{1.5} = 31.62$

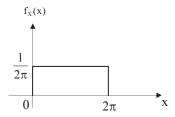
$$R_{max} = \left[\frac{P_t G A_e \times \sigma}{(4\pi)^2 \times S_{min}}\right]^{\frac{1}{4}} = \left[\frac{P_t G \sigma}{(4\pi)^2 \times \left(\frac{S_{min}}{A_e}\right)}\right]^{\frac{1}{4}} = \left[\frac{100 \times 10^3 \times 31.62 \times 200}{(4\pi)^2 \times 1 \times 10^{-9}}\right]^{\frac{1}{4}}$$

:. $R_{max} = 7955.05 \text{m} \text{ (or) } 7.9 \text{km} \text{ (or) } \cong 8 \text{km}$



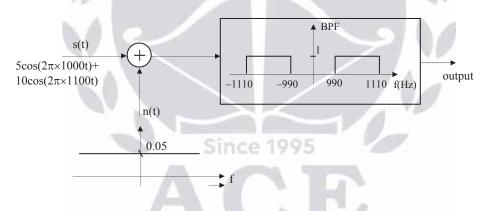
- 14. X is a random variable which is uniformly distributed between $(0, 2\pi)$. Then $E[\cos X] =$ _____.
- 14. Ans: 0

Sol:



$$E[\cos X] = \int_{0}^{2\pi} \cos x \ f_{X}(x) dx = \int_{0}^{2\pi} \cos x \frac{1}{2\pi} dx$$
$$= \frac{1}{2\pi} \sin x \Big|_{0}^{2\pi}$$
$$= 0$$

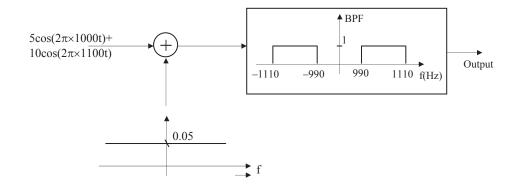
15. Consider the block diagram shown in figure. Find the signal to noise ratio at output.



15. Ans: 5.21

Range: 5.1 to 5.3

Sol:

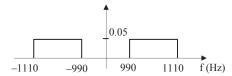




s(t) contains frequencies 1000 Hz, 1100 Hz. Since BPF allows from 990 Hz to 1110 Hz.

Output contains both signals, so output signal power = $\frac{5^2}{2} + \frac{10^2}{2} = 62.5$ W

Output Noise PSD is

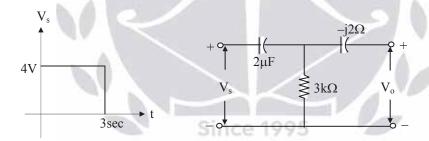


Area under PSD gives total power,

Output noise power = $2 \times 0.05 \times 120 = 12W$

Output signal to noise ratio = $\frac{62.5}{12}$ = 5.21

A square pulse of 4V amplitude is applied to RC circuit shown in figure. The capacitor is initially 16. uncharged. The output voltage V_0 at time $t = 3 \sec is$ (in Volts).



16. Ans: -4V

Sol:

At
$$t = 0$$
: C \rightarrow Short Circuit At $t = (3-)$: C \rightarrow Open Circuit

$$V_{C}(3-) = V_{C}(3+) = 4V$$

At t = 3: C \rightarrow Short Circuit with 4V source

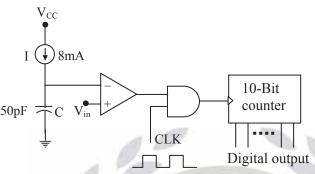
$$V_{c}(3) = -4V$$

Note: In the circuit $-j2\Omega$ capacitor Voltage is zero for all time since there is no current is flowing through that capacitor. Hence it acts as a Short Circuit.



A single slope ADC shown below is used to convert an analog input of 16V to digital. Determine the digital output (in decimal) if the clock frequency is 500 MHz. Initial voltage across

capacitor is 0V



17. Ans: 50

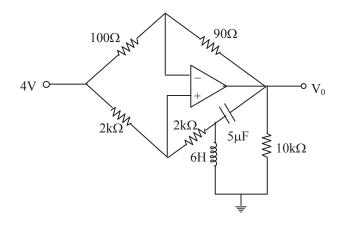
Sol:

Initially $V_c = 0V$; $V_{in} = 16V \Rightarrow$ comparator output = 1 & clock pulses reach the Counter

Time required for the capacitor to charge to 16V is It = CV

$$\Rightarrow t = \frac{CV}{I} = \frac{50 \times 10^{-12} \times 16}{8 \times 10^{-3}} = 100 \text{ns}$$
Clock period T = $\frac{1}{500 \times 10^{6}} = 2 \text{ ns}$
Digital output = $\frac{100 \text{ ns}}{T} = \frac{100 \text{ns}}{2 \text{ns}} = (50)_{10}$

If the Op-amp is ideal. Find the current (in Amp) across 90Ω resistor. 18.





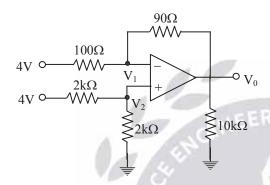
Ans: 0.02 Range: 0.01 to 0.03

Sol: Since, DC 4V is given as input, no switch (transient) is provided

 \Rightarrow we can proceed under steady state.

 $C \rightarrow$ behaves as open & $L \rightarrow$ Acts as short

the circuit is \rightarrow



Hence the circuit behaves as a subtractor.

:. From superposition,

$$V_0 = \frac{-90}{100} [4] + \left[1 + \frac{90}{100} \right] \left[\frac{4[2k]}{2k + 2k} \right] \text{Volts}$$

$$= -3.6 + 3.8$$

$$= 0.2 \text{ V}$$

Hence, from circuit, as
$$V_2 = \frac{4(2k)}{2k + 2k} = V_1$$

$$\therefore V_1 = 2V$$

$$I_{90\Omega} = \frac{V_1 - V_0}{90} = 0.02A$$

If A, B & C are n × n matrices and |A| = 2, |B| = 3 & |C| = 5 then the value of $|A^2BC^{-1}| = ?$

19. Ans: 2.4 Range: 2 to 3

Sol:
$$|A^2 B C^{-1}| = \frac{|A||A||B|}{|C|} = \frac{2 \times 2 \times 3}{5} = \frac{12}{5}$$

= 2.4



- What is the value of $\underset{x\to\infty}{\text{Lt}} \sqrt{(x^2+x+1)} x$
- 20. Ans: 0.5 Range: 0 to 1

Sol:
$$\underset{x \to \infty}{\text{Lt}} \sqrt{(x^2 + x + 1)} - x = \underset{x \to \infty}{\text{Lt}} \left[\sqrt{x^2 + x + 1} - x \right] \left[\frac{\sqrt{x^2 + x + 1} + x}{\sqrt{x^2 + x + 1} + x} \right]$$
$$= \frac{1 + 0}{\sqrt{1 + 0 + 0} + 1} = \frac{1}{2} = 0.5$$

- An Engineer applies the input $r(t) = 2\sin(t 1.55)$ to a chemical process and measures the output as $y(t) = 0.4\sin(t - 1.55)$. What is the gain of the system?
- 21. **Ans: 0.2** Range: 0.1 to 0.3
- **Sol:** Input $r(t) = 2\sin(t-1.55)$

Output $y(t) = 0.4\sin(t-1.55)$

For sinusoidal input, the output is

$$y(t) = A \times M \sin(t \pm \theta \pm \phi)$$

 $A \times M = 0.4$ (A \rightarrow Amplitude of input), A = 2

$$2M = 0.4 \Rightarrow M = 0.2$$

Find the difference equation relating input x(n) and output y(n) for the realization shown in 22. figure. If the filter impulse response is $h(n) = 0.5\delta(n) + 0.5\delta(n-1)$.

$$x(n)$$
 $+$
 \sum
filter
 z^{-1}

Option A:
$$y(n) + y(n-1) + y(n-2) = x(n) + x(n-1)$$

Option B:
$$y(n) + 0.5y(n-1) + 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

Option C:
$$y(n) - 0.5y(n-1) - 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

Option D:
$$1.5y(n) + 0.5y(n-1) = 0.5x(n) + 0.5x(n-1)$$



Ans: (B)

Sol: Assume filter transfer function is $H_1(z)$.

$$H_1(z) = 0.5 + 0.5z^{-1}$$

$$H(z) = \frac{Y(z)}{X(z)} = \frac{H_1(z)}{1 + z^{-1}H_1(z)} = \frac{0.5 + 0.5z^{-1}}{1 + z^{-1}(0.5 + 0.5z^{-1})}$$

$$\frac{Y(z)}{X(z)} = \frac{0.5 + 0.5z^{-1}}{1 + 0.5z^{-1} + 0.5z^{-2}}$$

$$Y(z)[1+0.5z^{-1}+0.5z^{-2}] = 0.5X(z) + 0.5z^{-1}X(z)$$

$$y(n) + 0.5y(n-1) + 0.5y(n-2) = 0.5x(n) + 0.5x(n-1)$$

Distractor Logic

Option A: If we miss 0.5 in filter impulse response.

Option B: Correct option

Option C: If we feel positive feedback system, wrong interpretation of taking negative coefficients.

Option D: In the feedback multiplier of $h_1(n)$ if we miss z^{-1}

An RC low pass filter has the impulse response $h(t) = e^{-t} u(t)$. The response of the system due 23. to the input $x(t) = e^{2t} u(-t)$ is _____

Option A:
$$\frac{1}{3}e^{2t}u(-t) + \frac{1}{3}e^{-t}u(t)$$

Option B:
$$\frac{-1}{3}e^{2t}u(t) - \frac{1}{3}e^{-t}u(-t)$$

Option C:
$$-\frac{1}{3}e^{2t}u(-t) - \frac{1}{3}e^{-t}u(t)$$

Option D:
$$e^{2t}u(-t) + e^{-t}u(t)$$

23. Ans: (A)

Sol:
$$H(s) = \frac{1}{s+1}$$
; $\sigma > -1$

$$X(s) = \frac{-1}{s-2}$$
; $\sigma < 2$

Output ROC =
$$(\sigma > -1) \cap (\sigma < 2) = -1 < \sigma < 2$$



$$Y(s) = X(s)H(s) = \frac{-1}{(s-2)(s+1)} = \frac{-1/3}{s-2} + \frac{1/3}{s+1}$$

Based on the output ROC, take inverse Laplace transform $y(t) = \frac{1}{2}e^{2t}u(-t) + \frac{1}{2}e^{-t}u(t)$

Distractor Logic

Option A: Correct Answer

Option B: In the partial fraction if we feel pole '2' is right sided & pole '-1' is left sided

Option C: In the partial fraction expansion if we take negative sign of Y(s) as it is

Option D: In the partial fraction expansion if we miss $\frac{1}{3}$ multipliers

The diode D shown in below figure is connected to 1 for a long time and is connected to 2 at t = 0. 24.

At t = 0+, the current (i) is . (Assume the forward drop of the diode to be 0V)

$$\begin{array}{c|c}
1 & D \\
\hline
2 & t = 0 \\
\hline
20V & i \end{array}$$

$$\begin{array}{c|c}
1 & 1 & 1 & 1 \\
\hline
1 & 20V & i \\
\hline
\end{array}$$

Option A: 0 A

Option B: 10 mA

Option C: -20 mA

Option D:-10 mA

24. Ans: (C)

Sol: The drop across the diode does not change for sudden change in applied voltage.

$$\Rightarrow$$
 At t = 0+, i = $\frac{-20}{1k\Omega}$ = -20mA

Distractor Logic

Option A: If the student assumes the diode reverse current is zero, he/she will go wrong.

Option B: If he/she assumes that diode current does not change, he/she will go wrong.

Option C: Correct option

Option D: If the diode forward current is just reversed, he/she will go wrong.



The electric field intensity is given inside a sphere of radius $R \le b$ m as $\vec{E} = 4R^2\hat{r}$ (N/C). If the 25. sphere has permittivity ' ϵ ', then the total electric displacement leaving the sphere $R = \frac{b}{2}$ m will be

Option A: 4πεb⁴ Coulomb

Option B: πεb⁴ Coulomb

Option C: πb⁴ Coulomb

Option D: $\frac{\pi b^4}{4}$ Coulomb

25. Ans: (B)

Sol: Given: $\vec{E} = 4R^2\hat{r}$ (N/C); $R \le b$

$$\vec{D} = \varepsilon \vec{E}$$

$$= 4 \varepsilon R^2 \hat{r} C/m^2$$

From Gauss's Law

$$\psi_{\text{net}} \equiv Q_{\text{enc}} = \oint\limits_{s} \vec{D}.d\vec{S}$$

$$ψ_{net} = D_r \times Area$$

$$= 4εR^2 \times 4πR^2$$

$$= 16πεR^4$$

The net electric flux leaving the sphere of radius $R = \frac{b}{2}$

Since

$$\psi_{net} = 16\pi\epsilon \left(\frac{b}{2}\right)^4$$

$$\therefore \psi_{net} = \pi \epsilon b^4 C$$

Distractor Logic

Option A: Simplification mistake

$$\psi_{net} = 16\pi\epsilon \left(\frac{b}{2}\right)^4 = 16\pi\epsilon \frac{b^4}{4} = 4\pi\epsilon b^4$$

Which is wrong answer



Option B: Correct option

Option C: If we take directly \vec{E} , while applying Gauss's law,

then
$$\psi_{net} = 4R^2 \times 4\pi R^2$$

$$=16\pi\left(\frac{b}{2}\right)^4=\pi b^4$$

Which is wrong answer

Option D: $\psi_{net} = D_r \times Area$

If we take area = πR^2 then which results $\psi_{net} = \frac{\pi b^4}{4}$, which is wrong answer.

NEW BATCHES FOR

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How deep does a radar wave at 3GHz propagate in seawater ($\sigma = 4$ S/m, $\epsilon = 24\epsilon_0$, $\mu = \mu_0$) before 26. its amplitude is reduced to 10^{-6} of its amplitude just below the surface?

Option A: 0.459cm

Option B: 3.41cm

Option C: 4.08cm

Option D: 9.86cm

26. Ans: (D)

Sol:
$$E_0 e^{-\alpha z} = 10^{-6} \times E_0$$

$$z = \frac{\ln 10^6}{\alpha}$$

$$\frac{\sigma}{\omega \varepsilon} = \frac{4}{2\pi \times 3 \times 10^9 \times 24 \times \frac{10^{-9}}{36\pi}} = 1$$

$$\alpha = \omega \sqrt{\frac{\mu \epsilon}{2} \left(\sqrt{1 + \frac{\sigma^2}{\omega^2 \epsilon^2}} - 1 \right)}$$

$$=\frac{2\pi \times 3 \times 10^9}{3 \times 10^8} \sqrt{\frac{24}{2} \left(\sqrt{2}-1\right)}$$

$$\alpha = 140.04 \text{Np/m}$$

$$z = \frac{\ln 10^6}{140.04} = 0.0986m$$

$$\therefore$$
 z = 9.86cm

Distractor Logic

Option A: If we calculate skin depth,

$$\delta = \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{2}{2\pi\times3\times10^{9}\times4\pi\times10^{-7}\times4}}$$

= 0.459 cm

Which is wrong answer.

Option B: If we forgot to substitute ' ε_r ' value in α ,



$$\alpha = \frac{2\pi \times 3 \times 10^9}{3 \times 10^8} \sqrt{(\sqrt{2} - 1)}$$
 which gives us z = 3.41cm, hence incorrect answer

Option C:
$$\alpha = \omega \sqrt{\frac{\mu\epsilon}{2} \left(\sqrt{1 + \frac{\sigma^2}{\omega^2 \epsilon^2}} + 1 \right)}$$
 \rightarrow which is incorrect formulae.

$$= 338.17$$

$$z = \frac{\ln 10^6}{338.17} = 4.08$$
cm, which is wrong option

Option D: correct option.

Which of the following could not be the auto correlation function of a random process? 27.

Option A:
$$R(\tau) = \begin{cases} 1 - |\tau|; & |\tau| < 1 \\ 0; & |\tau| > 1 \end{cases}$$

Option B:
$$R(\tau) = 5\sin 3\tau$$

Option C:
$$R(\tau) = \begin{cases} \cos \tau; & |\tau| \le \pi/2 \\ 0; & |\tau| > \pi/2 \end{cases}$$

Option D:
$$R(\tau) = \frac{\sin \tau}{\tau}$$

27. Ans: (B)

Sol: Auto correlation function is always even function, maximum value occurs at $\tau = 0$. option (B) violating these two conditions

Since 1995

Distractor Logic

Option A: ACF satisfies $R_x(-\tau) = R_x(\tau)$

$$|R_x(\tau)| \le R_x(0)$$

$$R(\tau) = \begin{cases} 1 - |\tau|; & |\tau| < 1 \\ 0; & |\tau| > 1 \end{cases}$$
 satisfies both properties

So option A wrong

Option B: $R_x(-\tau) \neq R_x(\tau)$, $5\sin(3\tau)$ violates even condition So option (B) Correct Answer



Option C: $\cos(\tau)$, $|\tau| \le \pi/2$ satisfies all properties

So option C is wrong

Option D: $\frac{\sin \tau}{\tau}$ satisfies all properties

So option D is wrongs

A network is composed of two sub-networks N₁ & N₂ as shown in figure 28.



If the sub network N₁ contains only linear, bilateral, time invariant elements then it can be replaced by its Thevenin's equivalent even if the sub-network N₂ contains

Option A: A two-terminal element which is non linear

Option B: A non-linear inductance mutually coupled to an element in N₁

Option C: An element which is linear, but mutually coupled to same element in N₁

Option D: A dependent source the value of which depends upon the voltage (OR) current in any Since 1995 element of N₁

28. Ans: (A)

Sol: Network N₁ is passive network as per given data. Then network N₂ should be active element which should be independent. i.e., independent either voltage (OR) current source

Distractor Logic

Option A: Correct option

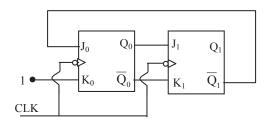
Option B: Nonlinear inductance is not possible

Option C: If N_2 is linear element we can't replace the network by its Thevenin's equivalent

Option D: If N₂ contains dependent source then control variable should be in network N₂ only



Find the value of the following counter after 730 clock pulses. Initially $Q_0 = Q_1 = 0$.



Option A: $Q_0 Q_1 = 00$

Option B: $Q_0 Q_1 = 01$

Option C: $Q_0 Q_1 = 10$

Option D: $Q_0 Q_1 = 11$

29. Ans: (C)

Sol: It is a 2 bit synchronous counter. Given $J_0 = \overline{Q}_1$; $K_0 = 1$; $J_1 = Q_0$; K_1

and the second									
1	Present state			Flip Flop Inputs			Next state		10 ::
v All	Q_0	Q ₁	J_0	K_0	J_1	K_1	Q_0	Q_1	0 (
0	0	0	1	1	0	1	1	0	2
2	1	0	1	1	1	0	0	1	①
1	0	1	0	1	0	1	0	0	0

counting sequence is 00, 10, 01, 00,..... it is a Mod-3 counter.

Counter value after 730 clock pulses is same value of the counter after 1 pulse i.e., $Q_0Q_1 = 10$

Distractor Logic

Option A: It might be mistaken as 2-bit Johnson counter which is a 4:1 counter. Thus value after 730 pulses is same as initial value i.e., $Q_0 Q_1 = 00$

Option B: The counting sequence of the counter may be mistaken as 00, 01, 10, 00, Then its solution is taken as 01

Option C: Correct option

Option D: Mistakenly chosen as 11



In 8085 Microprocessor the Accumulator has a 2's complement number '11110100'. Determine the function of executing the following sequence of instructions RLC, RRC and RAR. Assume 'carry' flag is cleared initially.

- (A) It finds 1's complement representation of the given 2's complement number
- (B) It converts the given 2's complement number to corresponding sign magnitude representation
- (C) It divides the given 2's complement number by 2
- (D) It multiplies the given 2's complement number by 2

Option A: It finds 1's complement representation of the given 2's complement number

Option B: It converts the given 2's complement number to corresponding sign magnitude representation

Option C: It divides the given 2's complement number by 2

Option D: It multiplies the given 2's complement number by 2

30. Ans: (C)

Sol: Given Accumulator value = -12_{10}

Accumulator value is 11111010 =

Distractor Logic

Option A: It may be mistaken as 1's complement form of given number

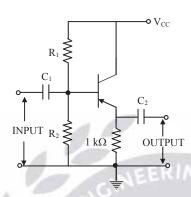
Option B: It may be mistaken as sign magnitude form of given number

Option C: Correct option

Option D: It may be mistaken as multiplication of 2's complement number



In the amplifier circuit shown in figure, the transistor parameters with usual notations are $g_m = 0.015~S,~r_{b'e} = 1~k\Omega~,~r_{bb'} = 90~\Omega~,~C_{b'e} = 20~pF~and~C_{b'c} = 3~pF~.~Neglecting~the~loading~effect~$ of biasing resistors, R₁ & R₂, the mid-frequency voltage gain of the amplifier is _____.



(A) - 13.76

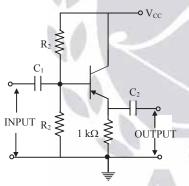
(B) 0.9411

(C) 0.936

(D) 1

31. Ans: (C)

Sol:



Consider the parameters of BJT

$$r_{b'e}=1\,k\Omega\,,\ r_{bb'}=90\Omega$$
 & $g_m=0.015$ S,

Step(1):
$$h_{ie} = r_{b'e} + r_{bb'} = 1090 \Omega$$

Consider,
$$r_{b'e} = \frac{h_{fe}}{g_m} \Rightarrow h_{fe} = r_{b'e} \times g_m = 1 \text{ k}\Omega \times 0.015 \text{ C} = 15$$

Step(2): The given circuit is emitter follower (CC Amplifier)

:. The mid-frequency voltage gain in a CC amplifier,

$$A_{_{\rm V}} = \frac{\left(1 + h_{_{\rm fe}}\right)\!R_{_{\rm E}}}{h_{_{\rm ie}} + \left(1 + h_{_{\rm fe}}\right)\!R_{_{\rm E}}} = \frac{16\!\times\!1\,{\rm k}\Omega}{1.09{\rm k}\Omega + 16\!\times\!1{\rm k}\Omega} = \frac{16\,{\rm k}\Omega}{17.09\,{\rm k}\Omega}$$

$$A_V = 0.936$$



Distractor Logic

Option: A: If the given circuit is assumed as CE Amplifier, with a load resistance of

$$R_L=1k$$
; then,

Step (1):
$$h_{ie} = r_{b'e} + r_{bb'} = 1090 \Omega$$

Consider,
$$r_{b'e} = \frac{h_{fe}}{g_m} \Rightarrow h_{fe} = r_{b'e} \times g_m = 1 \text{ k}\Omega \times 0.015 \text{ U} = 15$$

Step (2):
$$A_V = -\frac{h_{fe}R_L}{h_{ie}} = -\frac{15 \times 1k}{1.09k} = -13.76$$

Option: B

CC Amplifier: if we consider $h_{ie} = r_{b'e}$ [neglecting $r_{bb'}$ (base spreading resistance)]

$$A_{V} = \frac{\left(1 + h_{fe}\right)\!R_{E}}{h_{ie} + \left(1 + h_{fe}\right)\!R_{E}} = \frac{16 \times 1k\Omega}{1k\Omega + 16 \times 1k\Omega} = \frac{16k\Omega}{17k\Omega} = 0.9411$$

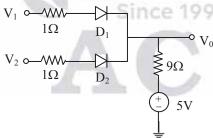
Option: C

CC Amplifier; $A_V = 0.936$

Option: D

The voltage gain in a CC amplifier ideally is '1'

Assume both the diodes to be precise, if $V_1 = -5V \& V_2 = 10V$, find V_0 ? 32.



Option A: -4V

Option B: 8.87 V

Option C: 9.5 V

Option D: -4.63 V

32. Ans: (C)

Sol: Assume D_1 OFF & D_2 ON,



$$\therefore V_0 = \frac{\left(\frac{10}{1}\right) + \left(\frac{5}{9}\right)}{\left(\frac{1}{1}\right) + \left(\frac{1}{9}\right)} = 9.5V$$

Now, check Assumption:

$$i_{D_2} = \frac{10 - 9.5}{1} = 0.5A > 0,$$

$$\therefore$$
 D₂ \rightarrow ON \rightarrow True.

$$V_{D_1} = V_1 - V_0 = -5 - 9.5 = -14.5V < 0$$

$$\therefore$$
 D₁ \rightarrow OFF is True

Distractor Logic

Option A: $-4 \text{ V} \rightarrow \text{This is possible only when D}_1 \rightarrow \text{short \& D}_2 \rightarrow \text{open}$ But from given biasing, this result is impossible

Option B: 8.87 V \rightarrow is possible for $D_1 \rightarrow$ OFF & $D_2 \rightarrow$ ON, but if diodes are practical.

But given diodes are precise ⇒ They are ideal

Option C: True

Option D: $-4.63 \text{ V} \rightarrow \text{This}$ is possible only when $D_1 \rightarrow \text{short}$, $D_2 \rightarrow \text{open } \& \text{ diodes are practical}$

33. The general solution of
$$\frac{dy^4}{dx^4} - 6\frac{dy^3}{dx^3} + 12\frac{dy^2}{dx^2} - 8\frac{dy}{dx} = 0$$
 is

(A)
$$y = C_1 + (C_2 + C_3 x + C_4 x^2) e^{2x}$$

(B)
$$y = (C_1 + C_2 x + C_3 x^2) e^{2x}$$

(C)
$$y = (C_1 + C_2 x + C_3 x^2 + C_4 x^3) e^{2x}$$

(D)
$$y = C_1 + C_2 x + C_3 x^2 + C_4 e^{2x}$$

33. Ans: (A)

Sol: The given equation is $(D^4 - 6D^3 + 12D^2 - 8D) y = 0$

$$D(D^3 - 6D^2 + 12D - 8) y = 0$$

$$D(D-2)^3=0$$

$$\therefore$$
 D = 0, 2, 2, 2

... The required solution is (A)



34.
$$L\{e^{-2t}[1-u(t-1)]\} = ?$$

(A)
$$\frac{1 - e^{(s+2)}}{(s+2)}$$

(B)
$$\frac{1 + e^{(s+2)}}{(s+2)}$$

(C)
$$\frac{1 - e^{-(s+2)}}{(s+2)}$$

(D)
$$\frac{1 + e^{-(s+2)}}{(s+2)}$$

34. Ans: (C)

Sol:
$$L\{e^{-2t}[1-u(t-1)]\} = L\{e^{-2t}g(t)\}$$
 (Where $g(t) = [1-u(t-1)]$)

(Where
$$g(t) = [1 - u(t-1)]$$
)

=
$$G(s+2)$$
 $\left(G(s) = L\{g(t)\} = \frac{1}{s} - \frac{e^{-s}}{s}\right)$

$$=\frac{1-e^{-(s+2)}}{(s+2)}$$

OUR ESE 2016 TOP 10 RANKERS IN ALL STREAMS









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29 RANKS IN TOP 10 IN ESE-2016



The closed loop transfer function of a unity feedback system is $\frac{C(s)}{R(s)} = \left(\frac{30}{s^2 + 5s + 36}\right)$. The steady 35. state error due to a unit step input is

Option A: $\frac{36}{66}$

Option B: $\frac{1}{6}$

Option C: $\frac{36}{30}$

Option D: None

35. Ans: (B)

Sol: Given CLTF $\frac{C(s)}{R(s)} = \frac{30}{s^2 + 5s + 36}$

Get OLTF G(s) =
$$\frac{30}{s^2 + 5s + 6}$$
, H(s) = 1

Steady state error for unit step input $e_{ss} = \frac{1}{1+k} = \frac{1}{1+\frac{30}{1+\frac$

$$e_{ss} = \frac{6}{36} = \frac{1}{6}$$

Distractor Logic

Option A: If the given transfer function is considered as a OLTF then $e_{ss} = \frac{A}{1+k}$

$$e_{ss} = \frac{1}{1 + \frac{30}{36}} = \frac{36}{66}$$

Option B: Correct Option

Option C: If the given transfer function is considered as OLTF and taken $e_{ss} = \frac{A}{k} = \frac{1}{\left(\frac{30}{36}\right)} = \frac{36}{30}$

Option D: If considered OLTF is unstable then e_{ss} is none (or) ∞



Q.36 – Q.65 carry two marks each.

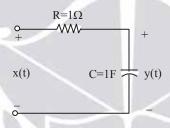
- Consider the filter $H(z) = \frac{z+2}{z+0.5}$. The input to this filter is $x(n) = \cos(n\pi)$. The phase delay of this system is _____ (in secs)
- 36. Ans: 1

Sol:
$$\omega_0 = \pi(z = e^{j\omega_0} = -1) \Rightarrow H(e^{j\omega_0}) = H(e^{j\pi}) = H(-1) = \frac{-1+2}{-1+0.5} = -2 = 2e^{-j\pi}$$

$$x(n) = \cos(n\pi) \Rightarrow y(n) = 2\cos(n\pi - \pi)$$

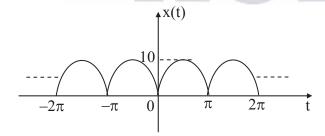
$$t_{p}(\omega) = -\frac{\theta(\omega)}{\omega} = -\frac{(-\pi)}{\pi} = 1 \sec$$

Signal $x(t) = |10\sin t|$ Volts is applied to the circuit shown in figure. The Fourier series coefficient of the output has a dc value of



Ans: 6.366

- Range: 6 to 7
- **Sol:** $x(t) = |10 \sin t|$ is a Full-wave rectifier wave form



$$T_0 = \pi$$

$$\omega_0 = \frac{2\pi}{T_0} = 2$$



The Exponential Fourier Series coefficients of Full Wave rectified wave form is

$$C_0 = \frac{1}{T_0} \int_0^{T_0} x(t) dt = \frac{1}{\pi} \int_0^{\pi} 10 \sin t dt = \frac{10}{\pi} \left[-\cos t \Big|_0^{\pi} \right]$$

$$C_0 = \frac{20}{\pi}$$

$$\begin{split} &C_n = \frac{1}{T_0} \int\limits_0^{T_0} x(t) e^{-jn\omega_0 t} dt = \frac{1}{\pi} \int\limits_0^{\pi} 10 \sin(t) e^{-jn2t} dt = \frac{10}{\pi} \int\limits_0^{\pi} \left[\frac{e^{jt} - e^{-jt}}{2j} \right] e^{-jn2t} dt \\ &= \frac{10}{2\pi j} \left[\int\limits_0^{\pi} e^{j(l-2n)t} dt - \int\limits_0^{\pi} e^{-j(l+2n)t} dt \right] \\ &= \frac{10}{2\pi j} \left[\frac{e^{j(l-2n)t}}{j(l-2n)} \Big|_0^{\pi} + \frac{e^{-j(l+2n)t}}{j(l+2n)} \Big|_0^{\pi} \right] \\ &= \frac{10}{2\pi j} \left[\frac{e^{j\pi(l-2n)} - 1}{j(l-2n)} + \frac{e^{-j\pi(l+2n)} - 1}{j(l+2n)} \right] = \frac{10}{2j\pi} \left[\frac{-2}{j(l-2n)} - \frac{2}{j(l+2n)} \right] \\ &= \frac{-20}{-2\pi} \left[\frac{1}{l-2n} + \frac{1}{l+2n} \right] \\ &C_n = \frac{10}{\pi} \left[\frac{2}{l-4n^2} \right] = \frac{20}{\pi(l-4n^2)} \end{split}$$

$$C_n = \frac{10}{\pi} \left[\frac{2}{1 - 4n^2} \right] = \frac{20}{\pi (1 - 4n^2)}$$

$$H(\omega) = \frac{1}{1+j\omega} \Rightarrow H(n\omega_0) = \frac{1}{1+jn\omega_0} = \frac{1}{1+j2n}$$

$$H(n\omega_0)\big|_{n=0}=1$$

$$\therefore$$
 dc component of output is = $C_0H(0) = \frac{20}{\pi} = 6.366$

- A MOS capacitor with substrate doping of 10¹⁹ atoms/cm³ of Boron and gate oxide thickness 1nm 38. is operated in strong accumulation. If the permittivity of oxide is 3.5×10⁻¹¹ F/m. The gate capacitance per unit area will be (mF/m^2) . (Given permittivity of Si = 1.05×10^{-10} F/m and the maximum depletion width in Si can be 20 nm).
- 38. **Ans: 35**

Sol:
$$C_{ox} = \frac{\varepsilon_{ox}}{t_{ox}} = \frac{3.5 \times 10^{-11}}{1 \times 10^{-9}} = 0.035 \text{ F/m}^2 = 35 \text{ mF/m}^2$$

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39. An n-type silicon sample with donor doping concentration $N_D = 10^{16}/\text{cm}^3$ is steadily illuminated such that there is an additional generation rate (g') of 10^{21} cm⁻³s⁻¹. If $\tau_{n0} = \tau_{p0} = 10^{-6}$ s, the position of quasi-Fermi level for holes with respect to intrinsic level is _____eV. (Assume intrinsic carrier concentration as 1.5×10¹⁰/cm³ and volt-equivalent temperature as 0.0259 V)

Since 1995

39. Ans: -0.2877

Range: -0.2 to -0.3

Sol: n-type

$$\Delta n = \Delta p = \ g' \ \tau_{po} = 10^{21} 10^{-6} = 10^{15} / cm^3$$

We have
$$n_0 = N_D = 10^{16} / \text{cm}^3$$

$$p_0 = \frac{n_i^2}{n_0} = \frac{(1.5 \times 10^{10})^2}{10^{16}} = 2.25 \times 10^4 / \text{cm}^3$$



$$E_{Fp} - E_{Fi} = -kT \ln \left(\frac{p_0 + \Delta p}{n_i} \right)$$
$$= -0.0259 \ln \left(\frac{2.25 \times 10^4 + 10^{15}}{1.5 \times 10^{10}} \right)$$
$$= -0.2877 \text{ eV}$$

A rectangular waveguide has a width to height ratio $\frac{a}{h} = 2$ and the ratio between operating frequency and the cutoff frequency is $\frac{f}{f_{r(x)}} = 2$ at f = 10GHz. What is the maximum timeaveraged power (in Mega Watts) that can be transmitted in the waveguide in the TE10 mode without exceeding the breakdown electric field intensity of 30kV/cm in air?

40. Ans: 2.3 Range: 2.1 to 2.5

Sol: Given $\frac{a}{h} = 2$, $E_0 = 30 \text{kV/cm} = 3 \times 10^6 \text{ V/m}$

$$\frac{f}{f_{c(10)}} = 2$$

$$f = 10GHz$$

$$f_{c}=\frac{f}{2}$$

$$\frac{c}{2a} = 5GHz$$

$$\Rightarrow \frac{3 \times 10^{10}}{2 \times a} = 5 \times 10^9$$

∴
$$a = 3cm$$

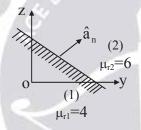
$$\eta_{\text{TE}_{10}} = \frac{\eta_0}{\sqrt{1 - \left(\frac{f_c}{f}\right)^2}} = \frac{120\pi}{\sqrt{1 - \left(\frac{1}{2}\right)^2}} = \frac{240\pi}{\sqrt{3}} \qquad \text{[:: for air filled } \eta_0 = 120\pi\Omega\text{]}$$



$$P_{total} = \frac{E_0^2}{\eta (TE_{10})} \frac{ab}{4}$$
$$= \frac{9 \times 10^{12} \times 9 \times 10^{-4}}{\left(\frac{240\pi}{\sqrt{3}}\right) 4 \times 2}$$

:.
$$P_{\text{total}} = 2.325 \times 10^6 \text{W} \text{ (or) } 2.32 \text{MW}$$

Region 1, with $\mu_{r1} = 4$, is the side of the plane y + z = 1 containing the origin (shown in figure). In region 2, $\mu_{r2} = 6$. If the magnetic flux density in region 1 is $\vec{B}_1 = 2\hat{a}_x + \hat{a}_y$ (Tesla), then the magnitude of magnetic flux density (in Tesla) in region 2 is



Ans: 3.2 41.

Range: 3 to 3.4

Since

Sol: The unit vector normal to the plane y + z = 1 is given by

$$\hat{\mathbf{a}}_{n} = \frac{\hat{\mathbf{a}}_{y} + \hat{\mathbf{a}}_{z}}{\sqrt{2}}$$

$$B_{n_1} = \vec{B}_1 \cdot \hat{a}_n = (2\hat{a}_x + \hat{a}_y) \cdot \left(\frac{\hat{a}_y + \hat{a}_z}{\sqrt{2}}\right)$$

$$B_{n1} = \frac{1}{\sqrt{2}}$$

$$\vec{B}_{n_1} = B_{n_1} \hat{a}_n = \frac{1}{\sqrt{2}} \left(\frac{\hat{a}_y + \hat{a}_z}{\sqrt{2}} \right)$$

$$\vec{B}_{n_1} = 0.5\hat{a}_y + 0.5\hat{a}_z$$

$$\vec{B}_{n2} = \hat{B}_{n_1} = 0.5\hat{a}_y + 0.5\hat{a}_z$$



$$\begin{split} \vec{B}_{t_1} &= \vec{B}_1 - \vec{B}_{n_1} \\ &= \left(2\hat{a}_x + \hat{a}_y \right) - \left(0.5\hat{a}_y + 0.5\hat{a}_z \right) \\ \vec{B}_{t_1} &= 2\hat{a}_x + 0.5\hat{a}_y - 0.5\hat{a}_z \\ \frac{B_{t_1}}{\mu_1} &= \frac{B_{t_2}}{\mu_2} \\ \Rightarrow \vec{B}_{t_2} &= \left(\frac{\mu_2}{\mu_1} \right) \vec{B}_{t_1} = \left(\frac{3}{2} \right) \left[2\hat{a}_x + 0.5\hat{a}_y - 0.5\hat{a}_z \right] \\ \vec{B}_{t_2} &= 3\hat{a}_x + 0.75\hat{a}_y - 0.75\hat{a}_z \\ \vec{B}_2 &= 3\hat{a}_x + 1.25\hat{a}_y - 0.25\hat{a}_z \quad \left[\because \vec{B}_2 = \vec{B}_{t_2} + \vec{B}_{n_2} \right] \\ \therefore \left| \vec{B}_2 \right| &= \sqrt{(3)^2 + (1.25)^2 + (-0.25)^2} = 3.259 \text{ Tesla} \end{split}$$

- 42. Bandwidth of an Angle Modulated signal $10\cos(2\pi \times 10^7 t + 20\cos(1000\pi t))$ is (kHz).
- 42. Ans: 21

Sol:
$$\theta_i = 2\pi \times 10^7 t + 20\cos(1000\pi t)$$

$$\frac{d\theta_{i}}{dt} = 2\pi \times 10^{7} - 20 \times 1000\pi \sin(1000\pi t)$$

$$f_i = \frac{1}{2\pi} \frac{d\theta_i}{dt} = 10^7 - 10000 \sin(1000\pi t)$$

$$\Delta f = f_{i,max} - f_{c} = 10000 \, Hz$$

$$f_m = 500Hz$$

$$BW = 2\Delta f + 2f_m = 2 \times 10000 + 2 \times 500$$
$$= 21 \text{ kHz}$$

43. Ten different signals are to be Time Division Multiplexed and transmitted using PCM. Four of these signals have a maximum frequency of 10kHz, two of them have maximum frequency of 15kHz, two other signals have a maximum frequency of 5kHz and remaining signals have a maximum frequency of 20kHz. The value of bit rate (in Mbps), if signals sampled at Nyquist rate and samples are represented using 10 bits is _____



Ans: 2.4

Range: 2.3 to 2.5

Sol: $r_{b,TDM} = n f_{s,TDM}$

$$n = 10$$

$$f_{s,TDM} = 4f_{s_1} + 2f_{s_2} + 2f_{s_3} + 2f_{s_4}$$

$$W_1 = 10k$$
, $f_{s_1} = 20k$

$$W_2 = 15k$$
, $f_{s_2} = 30k$

$$W_3 = 5k$$
, $f_{s_3} = 10k$

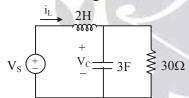
$$W_4 = 20k$$
, $f_{s_4} = 40k$

$$f_{s,TDM} = 4 \times 20k + 2 \times 30k + 2 \times 10k + 2 \times 40k$$

= 240 k samples/sec

 $r_{b,TDM} = 10 \times 240 \, k \, samples \, per \, sec = 2400 \, kbps = 2.4 \, Mbps$

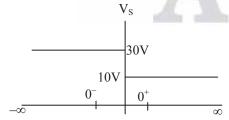
44. Consider the following network



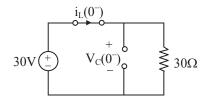
If $V_s = 10 + 20$ u(-t), then the value of $\frac{di_L(t)}{dt}$ at $t = 0^+$ in A/s is

44. Ans: -10

Sol:



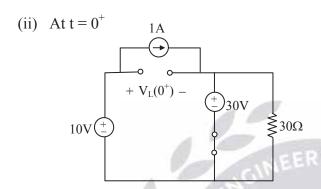
(i) At $t = 0^-$: Steady state





$$i_L(0^-) = \frac{30}{30} = 1A = i_L(0^+)$$

$$V_C(0^-) = 30V = V_C(0^+)$$



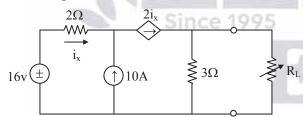
At $t = 0^+$: Transient state

By KVL
$$\Rightarrow 10 - V_L(0^+) - 30 = 0$$

$$\Rightarrow V_L(t)\Big|_{t=0^+} = -20$$

$$\Rightarrow \frac{Ldi_L(t)}{dt}\Big|_{t=0^+} = -20 \Rightarrow \frac{di_L(t)}{dt}\Big|_{t=0^+} = \frac{-20}{L} = \frac{-20}{2} = -10A/s$$

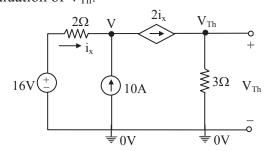
Consider the following network 45.



The value of maximum power transferred to load 'R_L' (in watts) is _____

Ans: 300 **45.**

Sol: Evaluation of V_{Th} :





Nodal
$$\Rightarrow$$
 $-i_x - 10 + 2i_x = 0 \Rightarrow i_x = 10A$

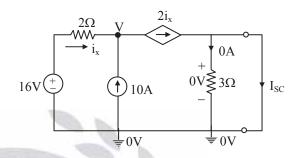
Nodal
$$\Rightarrow$$
 $-2i_x + \frac{V_{Th}}{3} = 0 \Rightarrow V_{Th} = 6i_x = 60V$

Evaluation of I_{SC}:

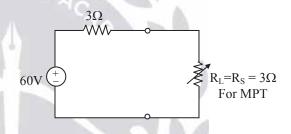
Nodal
$$\Rightarrow$$
 $-i_x - 10 + 2i_x = 0 \Rightarrow i_x = 10 A$

So,
$$I_{SC} = 2i_x = 20A$$

$$\Rightarrow R_{Th} = \frac{V_{Th}}{I_{SC}} = \frac{60}{20} = 3\Omega$$



$$\Rightarrow P_{\text{max}} = \frac{V_{\text{Th}}^2}{4 \times R_{\text{Th}}} = \frac{60^2}{4 \times 3} = 300 \text{W}$$



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06

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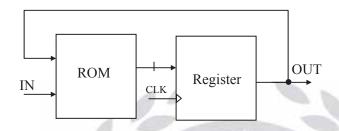


46. The maximum frequency of operation (in MHz) for the sequential circuit shown below is _____

The Propagation delays are as follows

$$t_{ROM} = 7.5 \text{ ns}; \ t_{Reg} = 2.5 \text{ ns}$$

$$t_{Setup} = 2.5 \text{ ns}; t_{Hold} = 2.5 \text{ ns}$$



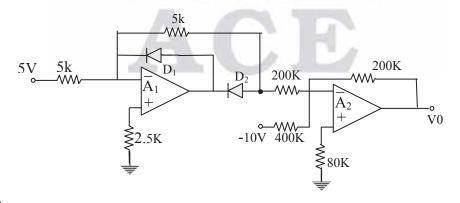
46. Ans: 80

Sol: The frequency of operation of the given sequential circuit is

$$f \le \frac{1}{t_{Rom} + t_{Reg} + t_{Setup}} \Rightarrow f \le \frac{1}{(7.5 + 2.5 + 2.5) \times 10^{-9}}$$

$$f_{\text{max}} = \frac{1 \times 10^9}{12.5} = 80 \text{MHz}$$

47. Find the output voltage V_0 (in Volts) of op-amp circuit shown in figure assuming op-amp & diodes are as ideal.



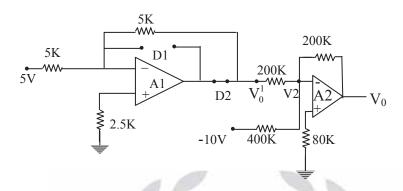
47. Ans: 10

Sol: Step (1) For the given input of 5v, D1 is off & D2 is ON

$$V_0^1 = \frac{-5k}{5k} \times 5v = -5v$$



Step (2) KCL at the inverting input, V₂ of op-amp2(A₂)



Op-amp 2 (A2)

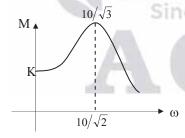
$$\frac{V_0^1}{200k} - \frac{10v}{400k} + \frac{V_0}{200k} = 0$$

$$V_0 = 200k \left[\frac{5V}{200k} + \frac{10V}{400k} \right] = 5 V + 5 V$$

$$\therefore V_0 = 10 \text{ V}$$

Frequency response of a second order system is given below. The value of K is ____ 48.

$$G(s) = \frac{K \times 100}{s^2 + 10s + 100}$$



48. Ans: 5

Sol:
$$G(s) = \frac{K \times 100}{s^2 + 10s + 100}$$

 $s^2 + 10s + 100 \Rightarrow s^2 + 2s\xi\omega_n + \omega_n^2$
 $\omega_n = 10$
 $\xi = 1/2$



$$M = \frac{K}{2\xi\sqrt{1-\xi^2}} = \frac{10}{\sqrt{3}}$$
$$= \frac{K}{2(0.5)\sqrt{1-(0.5)^2}} = \frac{10}{\sqrt{3}}$$

$$K = 5$$

49. $y = e^{-x} \left(C_1 \cos \sqrt{3} x + C_2 \sin \sqrt{3} x \right) + C_3 e^{2x}$ is the general solution of

(A)
$$(D^3 + 4) y = 0$$

(B) $(D^3 - 8) y = 0$
(C) $(D^3 + 8) y = 0$
(D) $(D^3 - 2D^2 + D)$

(B)
$$(D^3 - 8) y = 0$$

(C)
$$(D^3 + 8) y = 0$$

(D)
$$(D^3 - 2D^2 + D - 2) y = 0$$

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49. Ans: (B)

Sol: The roots of AE are $\left(-1 \pm \sqrt{3} i\right) \& 2$

:. The required equation is

$$(D-2)(D^2+2D+4)y=0$$

i.e.,
$$(D^3 - 8) y = 0$$

Find the value of $\int_{C} \frac{z \cos z}{\left(z - \frac{\pi}{2}\right)^2} dz$, where 'C' is |z - 1| = 1

(A)
$$i\pi$$

(B)
$$-i\pi$$

(C)
$$i\pi^2$$

$$(D) -i\pi^2$$

50. Ans: (D)

Sol: $z = \frac{\pi}{2} = \frac{3.14}{2} = 1.57$ is a pole of order '2' lies inside 'C'

$$\iint_{C} \frac{z \cos z}{\left(z - \frac{\pi}{2}\right)^{2}} dz = 2\pi i f^{1}\left(\frac{\pi}{2}\right) \quad \text{(where } f(z) = z \cos z\text{)}$$

$$=2\pi i \left(\frac{-\pi}{2}\right) = -\pi^2 i$$



51.
$$(a\alpha + b) x + ay + bz = 0$$

$$(b\alpha + c) x + by + cz = 0$$

 $(a\alpha + b) y + (b\alpha + c) z = 0$ have non-trivial solutions if

- (i) a, b, c are in A. P
- (ii) a, b, c are in G. P
- (iii) a, b, c are in H. P
- (iv) ' α ' is a root of $(ax^2 + 2bx + c) = 0$
- (A) both (i) & (iv)
- (B) both (ii) & (iv)
- (C) both (iii) & (iv)
- (D) only (iv)

51. Ans: (B)

Sol:
$$\begin{vmatrix} (a\alpha + b) & a & b \\ (b\alpha + c) & b & c \\ 0 & (a\alpha + b) & (b\alpha + c) \end{vmatrix} = 0$$

$$R_3 \rightarrow R_3 - (\alpha R_1 + R_2)$$

$$\Rightarrow \begin{vmatrix} a\alpha + b & a & b \\ b\alpha + c & b & c \\ -a\alpha^2 - 2b\alpha - c & 0 & 0 \end{vmatrix} = 0$$

i.e.,
$$-(a\alpha^2 + 2b\alpha + c)(ac - b^2) = 0$$

∴ '
$$\alpha$$
' is a root of $(ax^2 + 2bx + c) = 0$
or a, b, c are in G. P

Which one of the following statement is NOT TRUE about eigen signal? 52.

Option A: Every signal is an eigen signal to the system described by $h(t) = A\delta(t)$

Option B: The signal $x(t) = e^{i\beta t}$ is an eigen signal of an LTI system given by $h(t) = e^{-\alpha t}u(t)$

Option C: The signal $x(t) = \cos \beta t$ is an eigen signal of an LTI system given by $\dot{y}(t) + \alpha y(t) = x(t)$

Option D: The signal $x(t) = \operatorname{Sinc}(\alpha t)$ is an eigen signal of LTI system having $h(t) = \operatorname{Sinc}(\beta t)$; $\beta \ge \alpha$

52. Ans: (C)

Sol: If
$$x(t) = \cos(\beta t)$$
 & $\dot{y}(t) + \alpha y(t) = x(t)$ then
$$y(t) = A\cos(\beta t) + B\sin(\beta t) = C\cos(\beta t + \theta)$$



Option A: $x(t)*A\delta(t) = Ax(t)$ but you may feel $A\delta(t)$

Option B: we may think of ' α ' & ' β ' nature

Option C: Correct option

Option D: Wrong option

Sinc(t)*Sinc(t) = Sinc(t)

 $\operatorname{Sinc}(t) * \operatorname{Sinc}(2t) = \frac{1}{2} \operatorname{Sinc}(t)$ but you may think $\operatorname{Sinc}(2t)$

53. An ideal transistor has an emitter efficiency of 0.999 and collector-base leakage current of 10μA.

The active region emitter current due to holes, if $I_B = 0$ is

Option A: 0 A

Option B: 10 µA

Option C: -10 µA

Option D: 10 mA

53. **Ans: (D)**

Sol: For ideal transistor $\alpha_0 = \gamma = 0.999$; $\beta_0 = \frac{\alpha_0}{1-\alpha_0}$

 I_{CB0} = 10 $\mu A.$ Therefore I_{CE0} = (1+ $\beta_0)$ I_{CB0}

$$= (1+999) 10 \times 10^{-6}$$

= 10 mA

Distractor Logic

Option A: I_B is given as zero

If student assumes the output current to be zero, he/she will go wrong

Option B: Since $I_B = 0$, if student assumes the reverse current to flow through emitter, he/she will go wrong.

Option C: If he/she assumes the reverse current to flow in opposite direction, they will go wrong

Option D: Correct option



If an n-type semiconductor is uniformly illuminated with light, producing a uniform excess generation rate G. Then find the change in semiconductor conductivity ($\Delta \sigma$). (Let μ_n = electron mobility, μ_p = hole mobility, n = electron concentration, p = hole concentration, τ_n = electron life time, τ_p = hole life time and q = electron charge).

Option A: $\Delta \sigma = qn\mu_n + qp\mu_p$

Option B: $\Delta \sigma = qn\mu_n + q\mu_p \tau_p G$

Option C: $\Delta \sigma = q \mu_n \tau_n G + q p \mu_p$

Option D: $\Delta \sigma = q \mu_n \tau_p G + q \mu_p \tau_p G$

54. Ans: (D)

Sol: $\sigma = q\mu_n n + q\mu_p p$; Before illumination $n_n = n_{no}$, $p_n = p_{no}$

after illumination, $n_n = n_{no} + \Delta n = n_{no} + \tau_p G$; $p_n = p_{no} + \Delta p = p_{no} + \tau_p G$

$$\Rightarrow \Delta \sigma = [q\mu_n(n_{no} + \Delta n) + q\mu_p(p_{no} + \Delta p)] - [q\mu_n n_{no} + q\mu_p p_{no}] = q(\mu_n + \mu_p)\tau_p G$$

Distractor Logic

Option A: This is the well known formula for conductivity. If a student just remembers formulas he/she will go wrong.

Option B: If a student assumes that there will be only excess minority carriers, he/she can go wrong

Option C: If a student assumes that there will be only excess majority carriers, he/she can go wrong

Option D: This is the correct option. There will be generation of both electrons and holes even if it is an n-type semiconductor.

If a plane wave of frequency 10MHz propagating in lake water ($\sigma = 4 \times 10^{-3}$ S/m $\epsilon_r = 81$ and 55. $\mu_r = 1$) and in seawater ($\sigma = 4$ S/m, $\varepsilon_r = 81$ and $\mu_r = 1$), then the skin depths in both lake water and seawater will be

Option A: 12m and 8cm respectively

Option B: 2.51m and 8 cm respectively

Option C: 8cm and 12m respectively

Option D: 12m and 1.19cm respectively



55. Ans: (A)

Sol:
$$\left[\frac{\sigma}{\omega \varepsilon}\right]_{\text{Lake}} = \frac{4 \times 10^{-3}}{2\pi \times 10 \times 10^{6} \times 81 \times \frac{10^{-9}}{36\pi}} = 8.88 \times 10^{-2} << 1$$

Hence at f = 10MHz, lake water behaves as good dielectric

skin depth,
$$\delta = \frac{1}{\alpha} = \frac{2}{\sigma} \sqrt{\frac{\epsilon}{\mu}}$$

$$= \frac{2}{4 \times 10^{-3}} \sqrt{81} \times \frac{1}{120\pi} = 11.9 \text{m} \approx 12 \text{m}$$

∴ $\delta \approx 12$ m (Lake water)

$$\left[\frac{\sigma}{\omega\epsilon}\right]_{Sea} = \frac{4}{2\pi \times 10 \times 10^{6} \times 81 \times \frac{10^{-9}}{36\pi}} = 88.8 >> 1$$

Hence at f = 10MHz, seawater behaves as good conductor.

skin depth,
$$\delta = \frac{1}{\alpha} = \sqrt{\frac{2}{\omega\mu\sigma}}$$

$$= \sqrt{\frac{2}{2\pi \times 10 \times 10^6 \times 4\pi \times 10^{-7} \times 4}}$$

$$\approx \frac{1}{4\pi} \, \text{m}$$

∴ $\delta \approx 8$ cm (seawater)

Distractor Logic

Option A: correct option

Option B: For seawater,
$$\delta = \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{2}{2\pi\times10\times10^6\times4\pi\times10^{-7}\times4}} = 8cm$$
 For lake water,
$$\delta = \sqrt{\frac{2}{\omega\mu\sigma}} = \sqrt{\frac{2}{2\pi\times10\times10^6\times4\pi\times10^{-7}\times4\times10^{-3}}} = 2.51m$$

Which is incorrect result

Option C: As per the information given, we need to calculate the skin depths in both lake water and sea-water respectively. But in this option order is reversed.



Option D: For lake water $\delta = \frac{2}{\sigma} \sqrt{\frac{\varepsilon}{\mu}}$

If we take for seawater
$$\delta = \frac{2}{\sigma} \sqrt{\frac{\epsilon}{\mu}} = \frac{2}{4} \sqrt{\frac{81\epsilon_0}{\mu_0}} = 1.19 \text{cm}$$

But which is incorrect answer

56. Parity check matrix of a particular (7, 4) code is $\begin{bmatrix} 0 & 1 & 0 & 1 & 1 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \end{bmatrix}$.

Then Generator matrix is

Option A:
$$\begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 0 & 1 \end{bmatrix}$$

Option B:
$$\begin{bmatrix} 1 & 1 & 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 1 & 0 & 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$

Option C:
$$\begin{bmatrix} 1 & 0 & 1 & 1 & 1 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix}$$

Option D:
$$\begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 0 & 1 \end{bmatrix}$$

56. Ans: (B)

$$n = 7$$
; $k = 4$

$$p = n - k = 7 - 4 = 3$$

$$H = \begin{bmatrix} I & P^T \end{bmatrix}$$



$$H = \begin{bmatrix} 1 & 0 & 0 & 1 & 0 & 1 & 1 \\ 0 & 1 & 0 & 1 & 1 & 1 & 0 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \end{bmatrix}$$

$$\mathbf{P}^{\mathrm{T}} = \begin{bmatrix} 1 & 0 & 1 & 1 \\ 1 & 1 & 1 & 0 \\ 0 & 1 & 1 & 1 \end{bmatrix}$$

$$P = \begin{bmatrix} 1 & 1 & 0 \\ 0 & 1 & 1 \\ 1 & 1 & 1 \\ 1 & 0 & 1 \end{bmatrix}$$

$$G = \begin{bmatrix} P_{k \times n - k} & I_{k \times k} \end{bmatrix} = \begin{bmatrix} 1 & 1 & 0 & 1 & 0 & 0 & 0 \\ 0 & 1 & 1 & 0 & 1 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 1 & 0 & 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$
Districted to a Logic

Option A:

$$H = [I \quad P^T]$$

$$G = [P \ I]$$
 ('.' correct formula)

$$G = \begin{bmatrix} I & P \end{bmatrix}$$
 ('.' I, P order wrong)

$$G = \begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 0 & 1 \end{bmatrix}$$

So option A wrong

Option B: Correct answer

Option C:
$$G = [P \ I]$$

$$G = \begin{bmatrix} 1 & 0 & 1 & 1 & 1 & 0 & 0 \\ 1 & 1 & 1 & 0 & 0 & 1 & 0 \\ 0 & 1 & 1 & 1 & 0 & 0 & 1 \end{bmatrix} \quad [\because P^{T} \text{ mis sin g, order of I wrong}]$$

So option C wrong



Option D:
$$G = \begin{bmatrix} I & P \end{bmatrix}$$

$$= \begin{bmatrix} 1 & 0 & 0 & 0 & 1 & 1 & 1 & 0 \\ 0 & 1 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 1 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 1 & 1 & 0 & 1 \end{bmatrix} \quad [\because \text{order of G wrong}]$$

So Option D wrong

A signal is given by $s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$. Find the minimum number of bits of quantization required so that the signal to quantization noise ratio is greater than 50dB.

Option A: 8

Option B: 7

Option C: 9

Option D: 10

57. Ans: (C)

Sol:
$$s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$$

Signal Power =
$$\frac{20^2}{2} + \frac{17^2}{2} = 344.5$$

 $10\log SQNR \ge 50 \text{ dB}$

$$SQNR \ge 10^5$$

$$SQNR \ge 100000$$

$$\frac{344.5}{QNP} \ge 100000$$

$$QNP \le \frac{344.5}{100000}$$

$$QNP \le 3.445 \times 10^{-3}$$

$$QNP = \frac{\Delta^2}{12} = \left(\frac{DR}{L}\right)^2 \times \frac{1}{12}$$

Dynamic range =
$$V_{max} - V_{min} = 37 - (-37) = 74$$

$$\left(\frac{74}{L}\right)^2 \times \frac{1}{12} \le 3.445 \times 10^{-3}$$



$$\frac{1}{L^2} \le \frac{3.445 \times 10^{-3} \times 12}{74 \times 74}$$

$$\frac{1}{L^2} \le 7.55 \times 10^{-6}$$

$$L^2 \ge 132450.331$$

$$L \ge 363.93$$

$$n \ge log_2 363.93$$

$$n \ge 8.51 \Rightarrow n \ge 9$$

Option A: $s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$

Signal Power = $\frac{20^2}{2}$ = 200 [: maximum amplitude signal consider)

 $10 \log SQNR \ge 50$

$$SQNR \ge 10^5$$

$$\frac{200}{\text{ONP}} \ge 100000$$

$$\frac{QNP}{200} \le \frac{1}{100000}$$

$$QNP \le \frac{200}{100000} \le \frac{1}{500} \Rightarrow \frac{\Delta^2}{12} \le \frac{1}{500} \Rightarrow \Delta^2 = 0.024 \Rightarrow \left(\frac{DR}{L}\right)^2 \le 0.024 \Rightarrow \left(\frac{40}{L}\right)^2 \le 0.024$$

$$\Rightarrow \left(\frac{L}{40}\right)^2 \le 41.67$$

$$L^2 \ge 66672$$

$$L \ge 258.21$$
, $n = 8$

So option A wrong

Option B: $s(t) = 20\cos(100\pi t) + 17\cos(500\pi t)$

Signal Power =
$$\frac{20^2}{2} + \frac{17^2}{2} = 344.5$$

$$SQNR \ge 10^5$$



$$\frac{344.5}{QNP} \ge 100000$$

 $QNP \le 0.003445$

$$\frac{\Delta^2}{12} \le 0.003445$$

$$\Delta^2 \le 0.04134$$

$$\left(\frac{DR}{L}\right)^2 \le 0.04134$$

$$\left(\frac{34}{L}\right)^2 \le 0.04134$$

(: dynamic range wrong)

$$L^2 \ge 34^2 \times \frac{1}{0.04134}$$

 $L^2 \ge 27963.23174$

 $L \ge 167.22$

 $n \ge 7.38$

So option B wrong

Option C: Correct answer

Option D: If dynamic range is 40, n = 10

So option D wrong

circuit R X_L RL series 5Ω and 5Ω , the applied voltage 58. $V(t) = 2\cos 3t + 4\sqrt{2}\cos(3t + 45^{\circ}) + 12\sin 3t$ Volts. Then find the resultant current RMS value of the circuit.

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Option A:
$$\frac{3}{\sqrt{5}}$$
 A

Option B: 1A

Option C: $\sqrt{2}$ A

Option D: None

58. Ans: (B)



Sol:
$$V(t) = 2\cos 3t + 4\sqrt{2}\cos(3t + 45^{\circ}) + 12\sin 3t$$

$$= 2\cos 3t + 4\sqrt{2}\left(\cos 3t \frac{1}{\sqrt{2}} - \sin 3t \frac{1}{\sqrt{2}}\right) + 12\sin 3t$$

$$= 6\cos 3t + 8\sin 3t$$

$$V_{RMS} = \sqrt{\frac{1}{2}(6^2 + 8^2)} = \sqrt{50} = 5\sqrt{2} \text{ Volts}$$

$$\left|I_{RMS}\right| = \left|\frac{V_{RMS}}{Z}\right| = \frac{5\sqrt{2}}{\sqrt{R^2 + X_L^2}} = \frac{5\sqrt{2}}{\sqrt{5^2 + 5^2}} = 1Amps$$

Option A:
$$V_{RMS} = \sqrt{\frac{1}{2}(2^2 + (4\sqrt{2})^2 + 12^2)} = \sqrt{90} = 3\sqrt{10} \text{ Volts}$$

$$\left|I_{RMS}\right| = \left|\frac{V_{RMS}}{Z}\right| = \frac{3\sqrt{10}}{\sqrt{50}} = \frac{3}{\sqrt{5}} Amps$$

Option B:
$$I_{RMS} = 1A$$

Option C:
$$V = 6\cos 3t + 8\sin 3t = 10\sin(3t + 36.86^{\circ}) = 10\angle 36.86^{\circ}$$

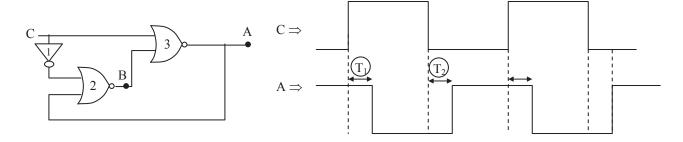
$$I = \frac{V}{Z} = \frac{10}{\sqrt{5^2 + 5^2}} = \frac{10}{\sqrt{50}} = \sqrt{2}Amps$$

Option D: None

In the following circuit 'C' input is driven by a square wave with 50% Duty cycle. 59.

Each gate has a propagation delay of 1ns. Determine the values of T₁ and T₂, respectively for the output waveform is as shown below.

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Option A: 3ns, 3ns

Option B: 1ns, 3ns

Option C: 1ns, 2ns

Option D: 2ns, 3ns

59. Ans: (B)

Sol: When C = 1, A becomes 0 after 1ns.

When C = 0, B becomes 0 after 2ns and A is 1 after 3ns

Distractor Logic

Option A: There is a possibility of adding propagation delays of all logic gates

Option B: Correct option

Option C: It is possible to take Gate-3 delay when C = 1 and propagation delays of Gate 2 and Gate \Im when C = 0

Option D: Possible to take Gate \mathbb{Q} , \mathbb{Q} delays when $\mathbb{C} = 1$ and Gate \mathbb{Q} , \mathbb{Q} and \mathbb{Q} delays when $\mathbb{C} = 0$

In a 12-bit bipolar ADC, the digital output is in 2's complement form. Find the digital output for 60. an analog input of -4V in 4 Hex digits.

Since 1995

The input voltage range is -4V to +4V

Option A: FFFF_H

Option B: 8000_H

Option C: F800_H

Option D: FF00_H

60. Ans: (C)

Sol: For -4V, the 12-bit digital output is $1000\ 0000\ 0000 = 800_{H}$.

Then digital output in 4 Hex digits is = $F800_H$

Distractor Logic

Option A: It is possible to select max negative output as

1111 1111 1111 which is FFFF_H in 4 Hex digits

Option B: Max negative value using 12-bits is 1000 0000 0000₂.

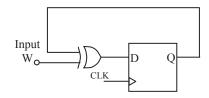
i.e., 800_H. In 4 Hex digits it may be mistaken as 8000_H

Option C: Correct option

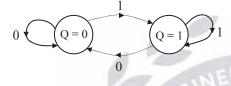
Option D: Mistakenly it is taken as F00_H, which is in 4 Hex digits as FF00_H



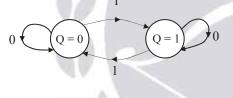
61. The state diagram for the sequential circuit shown below is



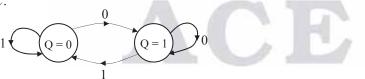
Option A:



Option B:

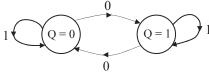


Option C:



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Option D:





Ans: (B)

Sol:
$$Q(t+1) = D$$
 and output $P = Q(t)$

Where
$$D = Q \oplus W$$

i.e.,
$$Q(t+1) = Q(t) \oplus W$$

If
$$W = 0 \Rightarrow Q(t+1) = Q(t)$$

If
$$W = 1 \Rightarrow Q(t+1) = \overline{Q(t)}$$

Distractor Logic

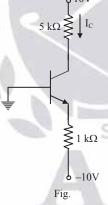
Option A: The output branches from each state are not assessed properly

Option B: Correct option

Option C: The output branches are not assessed properly

Option D: The output branches from each state are not assessed properly

In the circuit shown in figure , a silicon transistor with $V_{BE} = 0.7V$, $\beta = 100$ is used. Then find the 62. collector current I_C.



(A) 9.2 mA

(B) 2.1 mA

(C) 1.86 mA

(D) 0

62. Ans: (B)

Sol: Step (1): KVL for BE loop of BJT
$$0 - 0.7V - I_E 1K + 10V = 0 \dots (1)$$

$$\Rightarrow I_E = \frac{9.3V}{1K} = 9.3\text{mA} \dots (2) \text{ [i.e J}_E \text{ is FB]}$$

$$\Rightarrow I_C = \left(\frac{\beta}{1+\Omega}\right) I_E = 9.2\text{mA} \dots (3)$$



Step (2): KVL for C-loop

$$10V - I_C \times 5K - V_C = 0 \dots (4)$$

$$V_C = 10V - 9.2mA \times 5K = -36V \dots (5)$$

$$\Rightarrow$$
 V_{CB} = V_C - V_B = -36V - 0 = -36V (6)

NOTE: V_{CB} is -Ve, collector junctions is F.B

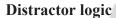
:. BJT is operated in saturation region

Step (3): \therefore BJT is in saturation, $V_{CE_{sat}} = 0.2V$

$$\Rightarrow$$
 $V_C = V_{CE_{sat}} + V_E = -0.5V \dots (1)$

KVL for collector –loop:

$$I_C = \frac{10V - (-0.5V)}{5K} = 2.1 \text{mA} \dots (2)$$



Option: A: If the device(BJT) is in forward active region, KVL for BE loop of BJT

$$0 - 0.7V - I_E 1K + 10V = 0$$

$$I_E = \frac{9.3 \text{V}}{1 \text{K}} = 9.3 \text{mA}$$

$$\therefore I_{c} = \left(\frac{\beta}{1+\beta}\right) I_{E} = \frac{100}{101} \times 9.3 \text{mA}$$

$$I_c = 9.2 \text{mA}$$

Option: B: $I_c = 2.1 \text{ mA} \dots (1)$ (Device is actually biased in saturation region)

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Option: C: If the device is in inverse (or reverse) active region,

(i.e) E-B junction is R.B & C.B junction is F.B

Assuming
$$V_{CB} = 0.7V \Rightarrow V_c = 0.7V \dots (1)$$

$$[\because V_{CB} = V_C - V_B = V_C]$$

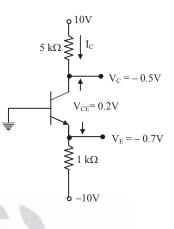
KVL for collector loop of BJT

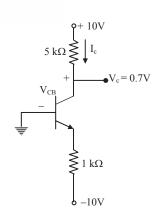
$$10V - I_c 5K - 0.7V = 0 \dots (1)$$

$$I_c = \frac{9.3 \text{V}}{5 \text{K}} = 1.86 \text{mA} \dots (2)$$

Option D: If the device is in cutoff region

$$I_B = 0 \Rightarrow I_C = 0 \dots (1)$$







- A bag contains (n + 1) coins. It is known that one of these coins shows heads on both sides where as the other coins are fair. One coin is selected at random and tossed. If the probability that the toss results in heads is $\frac{7}{12}$, then the value of 'n' is _____.
 - (A)4
- (B) 5
- (C) 6
- (D) 7

- Ans: (B) **63.**
- Sol: Let A = Event of selecting two headed coin
 - B = Event of selecting a fair coin
 - E = Event of coming head on selected coin

Given that
$$P(A) P(E/A) + P(B) \cdot P(E/B) = \frac{7}{12}$$

i.e.,
$$\frac{1}{(n+1)} \cdot 1 + \frac{n}{(n+1)} \cdot \frac{1}{2} = \frac{7}{12}$$

$$12 + 6n = 7n + 7$$

$$\therefore$$
 n = 5

The characteristic equation of a feedback control system is given by $s^2+s(k-1)+k=0$. Where 64. k > 0. The break points on RLD are

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Ans: (C)

Sol: Break point
$$\Rightarrow \frac{dk}{ds} = 0$$

Given, CE:
$$s^2 + sk - s + k = 0$$

$$CE \rightarrow (s^2-s)+k(s+1) = 0$$

$$k = -\frac{(s^2 - s)}{(s+1)}$$

$$\frac{dk}{ds} = -\left[\frac{(2s-1)(s+1) - (s^2 - s)}{(s+1)^2}\right] = 0$$

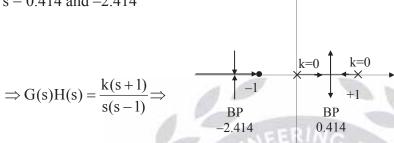


$$\Rightarrow 2s^2 - s + 2s - 1 - s^2 + s = 0$$

$$\Rightarrow$$
 s² + 2s - 1 = 0

$$s = \frac{-2 \pm \sqrt{4 - 4(1)(-1)}}{2} = \frac{-2 \pm \sqrt{8}}{2}$$

$$s = 0.414$$
 and -2.414



Option A: Given points are not on RLD.

Option B: One BP is valid and another is invalid.

Option C: Correct Option.

Option D: One BP is valid and another is invalid.

The state model of the system is given as 65.

$$\dot{\mathbf{X}} = \begin{bmatrix} 0 & 1 \\ -2 & -9 \end{bmatrix} \mathbf{X} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} \mathbf{U} & \mathbf{y} = \begin{bmatrix} 1 & 0 \end{bmatrix} \mathbf{x}$$

Then find the poles of the system.

Option A: s = -0.22, -8.77

Option B: s = 0.216, -9.216

Option C: s = 0.22, 8.77

Option D: s = -0.216, 9.21

65. Ans: (A)

Sol: From the magnitude equation

$$[\mathbf{sI} - \mathbf{A}] = \begin{bmatrix} \mathbf{s} & -1 \\ 2 & \mathbf{s} + 9 \end{bmatrix}$$



$$Adj[sI - A] = \begin{bmatrix} s + 9 & 1 \\ -2 & s \end{bmatrix}$$

$$TF = \frac{C Adj[sI - A]B}{|sI - A|} + p^{-0}$$

$$TF = \frac{\begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} s+9 & 1 \end{bmatrix} \begin{bmatrix} 0 \\ -2 & s \end{bmatrix} \begin{bmatrix} 1 \\ 1 \end{bmatrix}}{s(s+9)+2} = \frac{\begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} 1 \\ s \end{bmatrix}}{s^2+9s+2} = \frac{1}{s^2+9s+2}$$

CE
$$s^2 + 9s + 2 = 0$$

$$s = -0.22, -8.77$$

Option A: Correct option

Option B: This option is constructed calculation with wrong determinant $|sI - A| = s^2 + 9s - 2 = 0$

Option C: While calculating determinant, the subtraction of last row and column element is wrong, then the equation is $|sI - A| = s^2 - 9s + 2 = 0$

calculation Option D: This option is constructed with wrong determinant $as |sI - A| = s^2 - 9s - 2 = 0$

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